

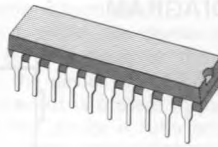


60W HI-FI DUAL AUDIO DRIVER

ADVANCE DATA

- WIDE SUPPLY VOLTAGE RANGE: 20 TO 90V (± 10 TO ± 45 V)
- VERY LOW DISTORTION
- AUTOMATIC QUIESCENT CURRENT CONTROL FOR THE POWER TRANSISTORS WITHOUT TEMPERATURE SENSE ELEMENTS
- OVERLOAD CURRENT PROTECTION FOR THE POWER TRANSISTORS
- MUTE/STAND-BY FUNCTIONS
- LOW POWER CONSUMPTION
- OUTPUT POWER 60W/8 Ω AND 100W/4 Ω

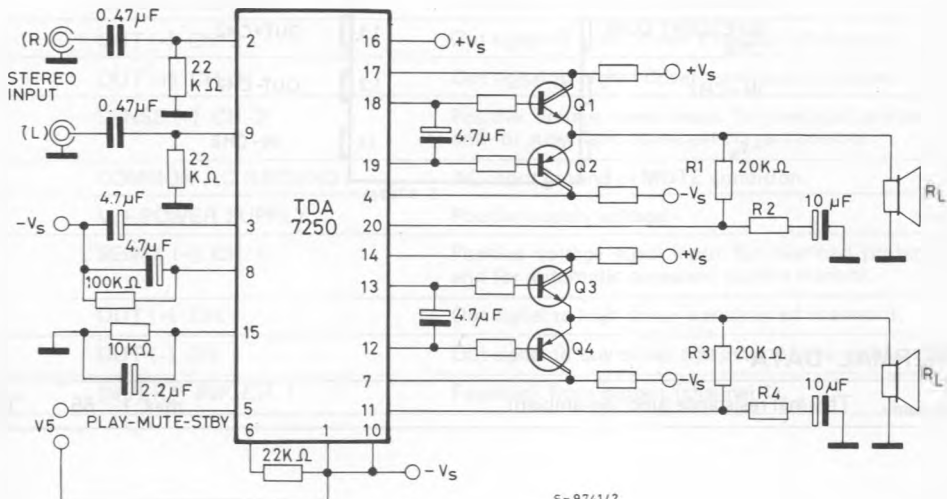
The TDA7250 stereo audio driver is designed to drive two pair of complementary output transistor in the Hi-Fi power amplifiers.



DIP-20 Plastic
(0.4)

ORDERING NUMBER: TDA7250

APPLICATION CIRCUIT



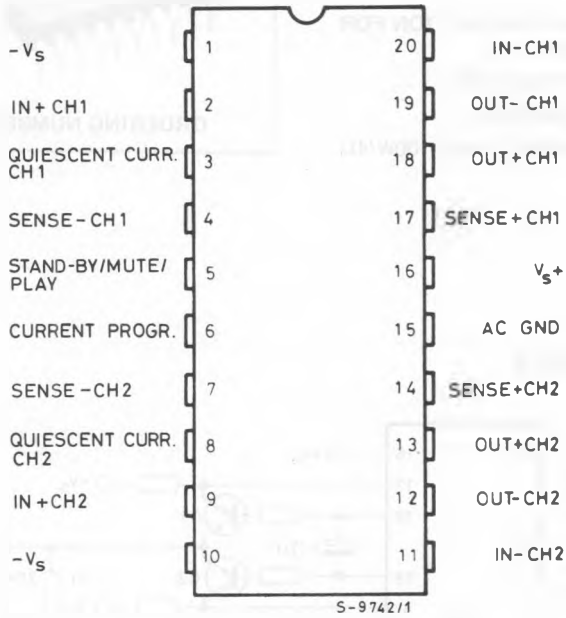
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ABSOLUTE MAXIMUM RATINGS

V_s	Supply voltage	100	V
P_{tot}	Power dissipation at $T_{amb} = 60^{\circ}\text{C}$	1.4	W
T_J, T_{stg}	Storage and junction temperature	-40 to +150	$^{\circ}\text{C}$

CONNECTION DIAGRAM

(Top view)



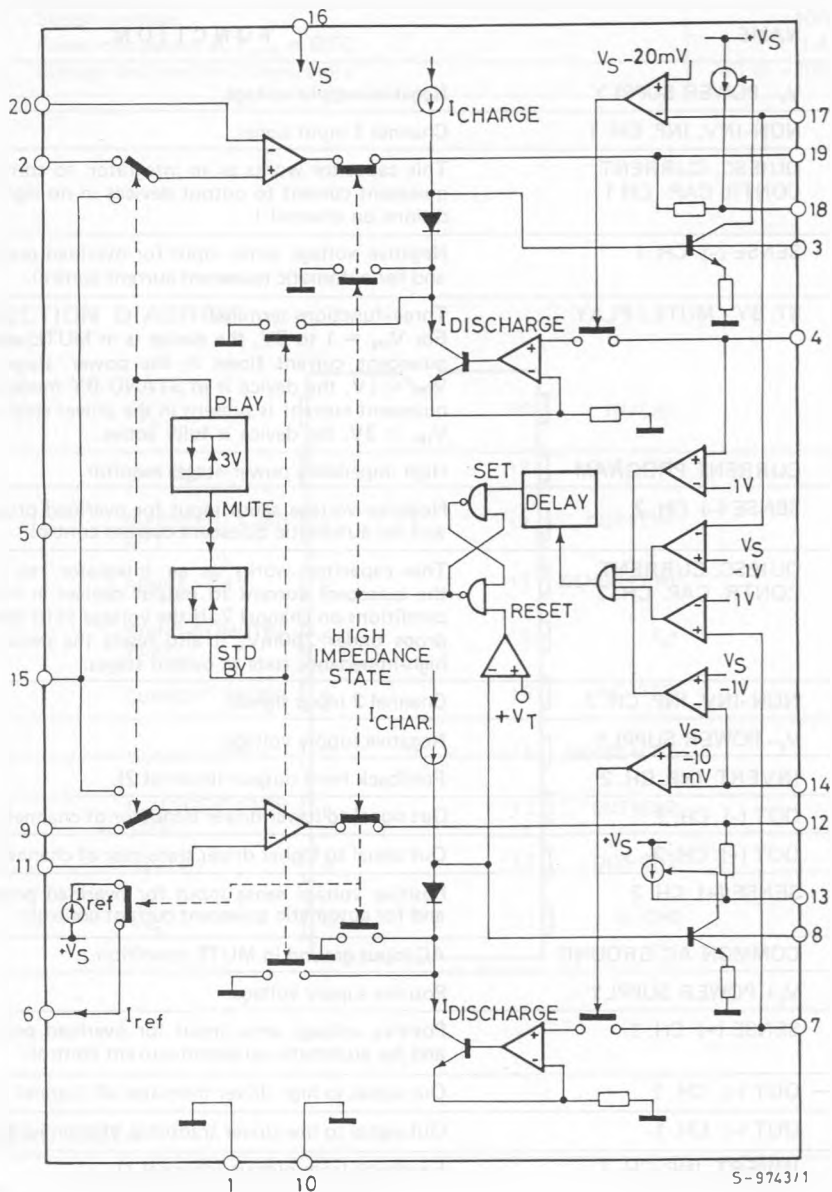
THERMAL DATA

$R_{th J-amb}$	Thermal resistance junction-ambient	max	65	$^{\circ}\text{C/W}$
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PIN FUNCTIONS

N°	NAME	FUNCTION
1	V _s - POWER SUPPLY	Negative supply voltage.
2	NON-INV. INP. CH. 1	Channel 1 input signal.
3	QUIESC. CURRENT CONTR. CAP. CH 1	This capacitor works as an integrator, to control the quiescent current to output devices in no-signal conditions on channel 1.
4	SENSE (-) CH. 1	Negative voltage sense input for overload protection and for automatic quiescent current control.
5	ST. BY / MUTE / PLAY	Three-functions terminal. For V _{IN} = 1 to 3V, the device is in MUTE and only quiescent current flows in the power stages;- for V _{IN} < 1V, the device is in STAND-BY mode and no quiescent current is present in the power stages; - for V _{IN} > 3V, the device is fully active.
6	CURRENT PROGRAM	High impedance power-stages monitor.
7	SENSE (-) CH. 2	Negative voltage sense input for overload protection and for automatic quiescent current control.
8	QUIESC. CURRENT CONTR. CAP. CH. 2	This capacitor works as an integrator, to control the quiescent current to output devices in no-signal conditions on channel 2. If the voltage at its terminals drops under 250mV, it also resets the device from high-impedance state of output stages.
9	NON-INV. INP. CH. 2	Channel 2 input signals.
10	V _s - POWER SUPPLY	Negative supply voltage.
11	INVERT. INP. CH. 2	Feedback from output (channel 2).
12	OUT (-) CH. 2	Out signal to lower driver transistor of channel 2.
13	OUT (+) CH. 2	Out signal to higher driver transistor of channel 2.
14	SENSE (+) CH. 2	Positive voltage sense input for overload protection and for automatic quiescent current control.
15	COMMON AC GROUND	AC input ground in MUTE condition.
16	V _s + POWER SUPPLY	Positive supply voltage.
17	SENSE (+) CH. 1	Positive voltage sense input for overload protection and for automatic quiescent current control.
18	OUT (+) CH. 1	Out signal to high driver transistor of channel 1.
19	OUT (-) CH. 1	Out signal to low driver transistor of channel 1.
20	INVERT. INP. CH. 1	Feedback from output (channel 1).

BLOCK DIAGRAM



ELECTRICAL CHARACTERISTICS ($T_{amb} = 25^{\circ}\text{C}$, $V_s = \pm 35\text{V}$, play mode, unless otherwise specified)

Parameter	Test Conditions		Min.	Typ.	Max.	Unit
V_s Supply voltage			± 10		± 45	V
I_d Quiescent drain current	Stand-by mode			8		mA
	Play mode			10	14	
I_b Input bias current				0.2	1	μA
V_{os} Input offset voltage				1	± 10	mV
I_{os} Input offset current				100	200	nA
G_v Open loop voltage gain	$f = 100\text{Hz}$			90		dB
	$f = 10\text{KHz}$			60		
e_N Input noise voltage	$R_G = 600\Omega$ $B = 20\text{Hz to } 20\text{KHz}$			3		μV
SR Slew rate				10		$\text{V}/\mu\text{s}$
d Total harmonic distortion	$G_v = 26\text{dB}$	$f = 1\text{KHz}$		0.004		%
	$P_O = 40\text{W}$	$f = 20\text{KHz}$		0.03		
V_{opp} Output voltage swing				60		V_{pp}
P_O Output power (*)	$V_s = \pm 35\text{V}$	$R_L = 8\Omega$		60		W
	$V_s = \pm 30\text{V}$	$R_L = 8\Omega$		40		
	$V_s = \pm 35\text{V}$	$R_L = 4\Omega$		100		
I_O Output current				± 5		mA
SVR Supply voltage rejection	$f = 100\text{Hz}$			75		dB
C_s Channel separation	$f = 1\text{KHz}$			75		dB

MUTE / STANDBY / PLAY FUNCTIONS

I_i Input current (pin 5)			0.1		μA
V_{th} Comparator standby/mute threshold (**)		1.0	1.25	1.5	V
H Hysteresis standby/mute			200		mV
V_{th} Comparator mute/play threshold (**)		2.4	3.0	3.6	V
H Hysteresis mute/play			300		mV
Mute attenuation	$f = 1\text{KHz}$		60		dB
V_i Input voltage max. (Pin 5)		12 (**)			V

 (*) Application circuit of fig. 1 $f = 1\text{KHz}$; $d = 0.1\%$; $G_v = 26\text{dB}$

 (**) Referred to $-V_s$

ELECTRICAL CHARACTERISTICS (continued)

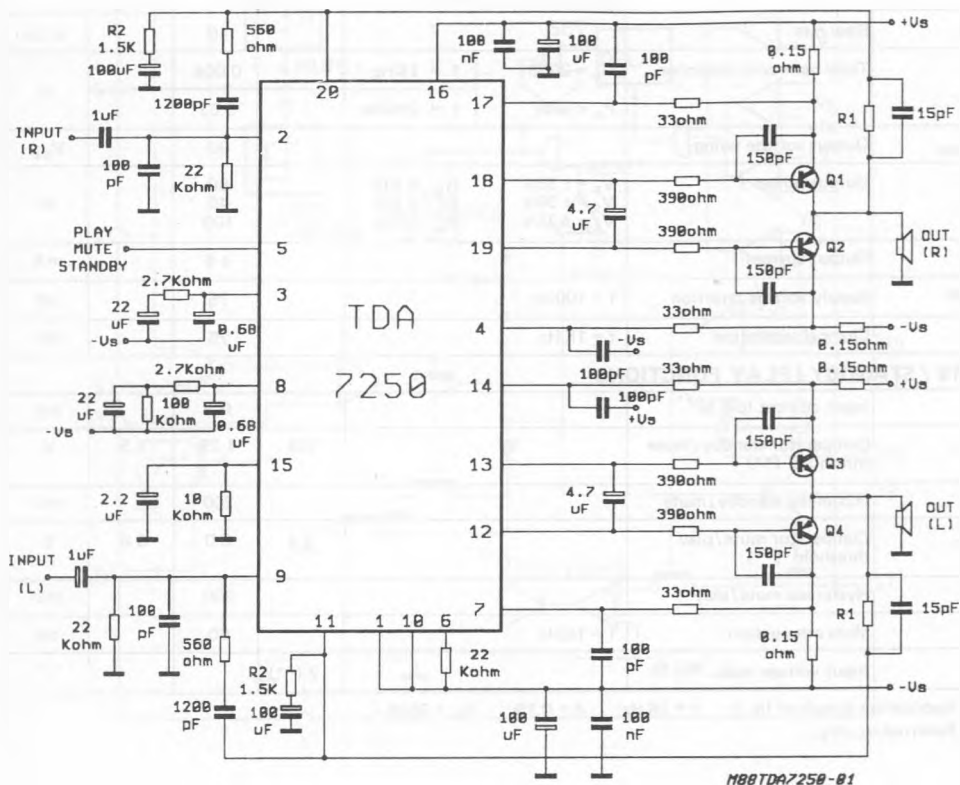
CURRENT SURVEY CIRCUITRY

Comparator reference	to $+V_S$ to $-V_S$	0.8 0.8	1 1	1.4 1.4	V V
t_d Delay time		10			μs

QUIESCENT CURRENT CONTROL

Capacitor current	Charge Discharge	30 250	60 500		μA μA
Comparator reference	to $+V_S$ to $-V_S$	10	20 10	25	mV mV

Fig. 1 - Application circuit with Power Darlings



NOTE: Q1/Q2 = Q3/Q4 = TIP 142/TIP 147
GV = $1 + R1/R2$

Fig. 2 - Output power vs. supply voltage

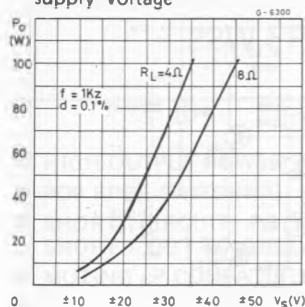


Fig. 3 - Distortion vs. output power (*)

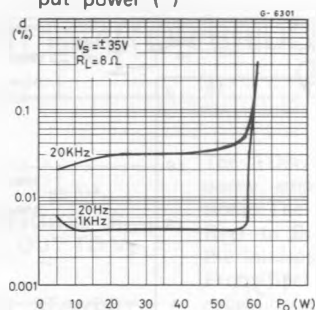


Fig. 4 - Channel separation

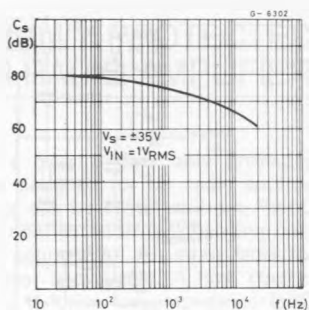


Fig. 5 - Supply voltage rejection vs. frequency

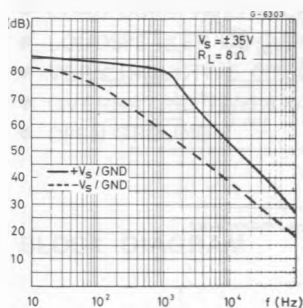


Fig. 6 - Quiescent current vs. supply voltage

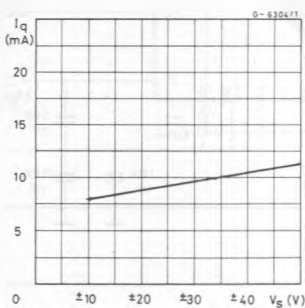


Fig. 7 - Quiescent current vs. Tamb

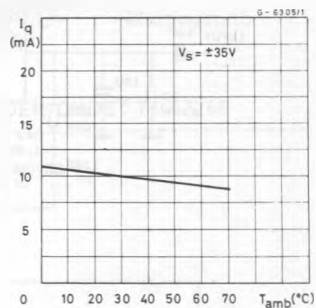


Fig. 8 - Total dissipated power vs. output power (*)

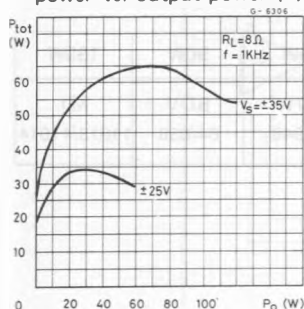


Fig. 9 - Efficiency vs. output power (*)

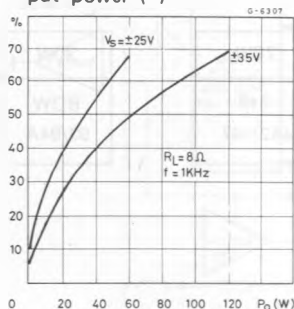
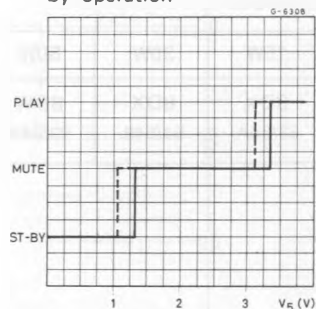


Fig. 10 - Play-mute stand-by operation



(*) Complete circuit

Fig. 11 - Application circuit using power transistors

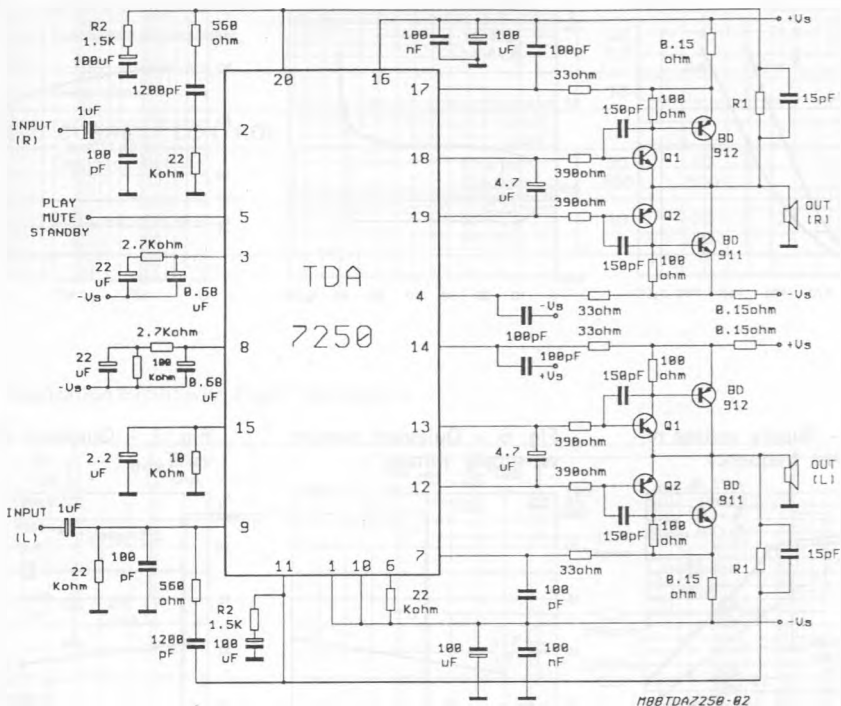


Fig. 12 – Suggested transistor types for various loads and powers.

$$R_L = 8 \Omega$$

15W	30W	50W	70W
BDX 53/54A	BDX 53/54B	BDW 93/94B	TIP 142/147

$$R_L = 4 \Omega$$

30W	50W	90W	130W
BDW 93/94A	BDW 93/94B	BDV 64/65B	MJ 11013/11014