SIEMENS

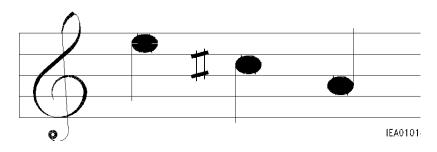
Programmable Single-/Dual-/Triple- Tone Gong

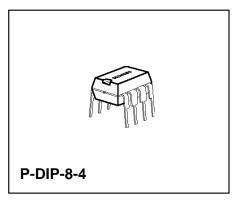
SAE 800

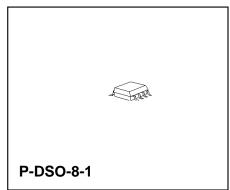
Preliminary Data Bipolar IC

Features

- Supply voltage range 2.8 V to 18 V
- Few external components (no electrolytic capacitor)
- 1 tone, 2 tones, 3 tones programmable
- Loudness control
- Typical standby current 1 μA
- Constant current output stage (no oscillation)
- High-efficiency power stage
- Short-circuit protection
- Thermal shutdown







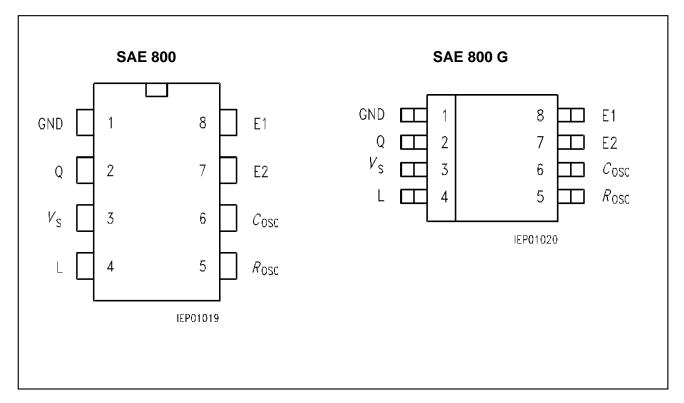
Туре		Ordering Code	Package		
\blacktriangledown	SAE 800	Q67000-A8339	P-DIP-8-4		
lacktriangledown	SAE 800 G	Q67000-A8340	P-DSO-8-1 (SMD)		

▼ New type

Functional Description

The SAE 800 is a single-tone, dual-tone or triple-tone gong IC designed for a very wide supply voltage range. If the oscillator is set to f_0 = 13.2 kHz for example, the IC will issue in **triple-tone-mode** the minor and major third $e^2 - C$ sharp - a, corresponding to 660 Hz - 550 Hz - 440 Hz, in **dual-tone-mode** the minor third $e^2 - C$ sharp, and in **single-tone-mode** the tone e^2 (derived from the fundamental frequency f_0 ; $f_1 = f_0 / 20$, $f_2 = f_0 / 24$, $f_3 = f_0 / 30$).

When it is not triggered, the IC is in a standby state and only draws a few μA . It comes in a compact P-DIP-8-1 or P-DSO-8-1 (SMD) package and only requires a few external components.



Pin Configuration

(top view)

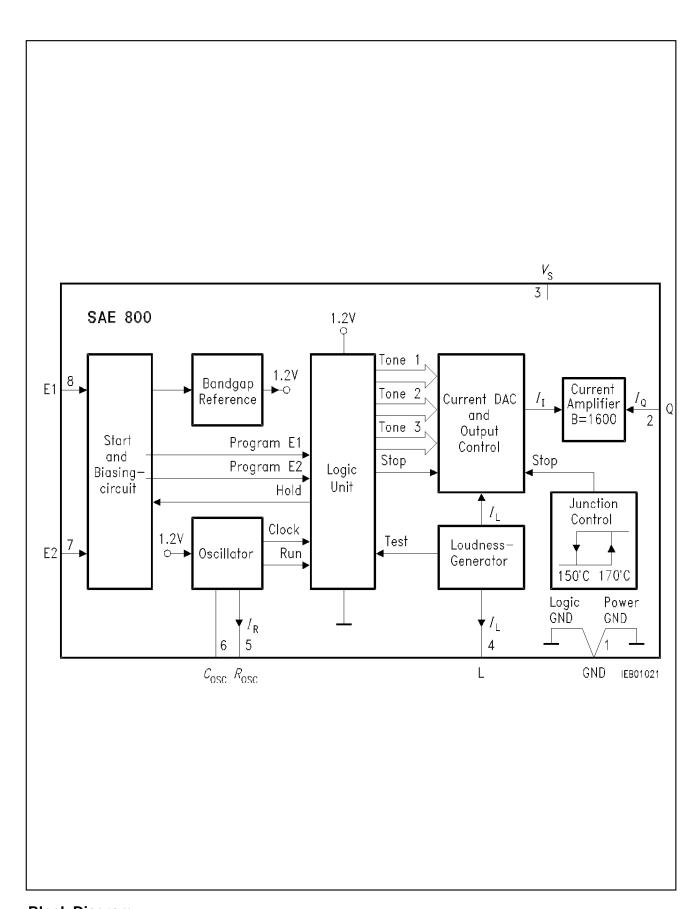
Pin Definitions and Functions

Pin	Symbol	Function
1	GND	Ground
2	Q	Output
3	V_{S}	Supply Voltage
4	L	Loudness Control
5	$R_{ extsf{OSC}}$	Oscillator Resistor
6	$C_{ m osc}$	Oscillator Capacitor
7	E2	Trigger 2 (dual tone)
8	E1	Trigger 1 (single tone)

Functional Description (cont'd)

An RC combination is needed to generate the fundamental frequency (pin $R_{\rm OSC}$, $C_{\rm OSC}$). The volume can be adjusted with another resistor (pin L). The loudspeaker must be connected directly between the output Q and the power supply $V_{\rm S}$. The current-sink principle combined with an integrated thermal shutdown (with hysteresis) makes the IC overload-protected and shortcircuit-protected.

There are two trigger pins (E1, E2) for setting single-tone, dual-tone or triple-tone mode.



Block Diagram

Circuit Description

Trigger

Positive pulses on inputs E1 and/or E2 trigger the IC. The hold feedback in the logic has a delay of several milliseconds. After this delay has elapsed, the tone sequence is started. This prevents parasitic spikes from producing any effect on the trigger pins.

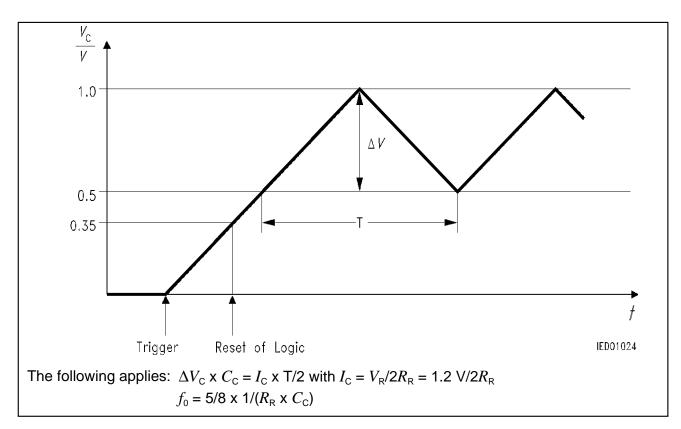
The following **table** shows the trigger options:

E1	E2	Mode	Issued Sequence
Triggered	Triggered	Triple-tone	Minor and major third
Grounded/open	Triggered	Dual-tone	Minor third
Triggered	Grounded/open	Single-tone	1st tone of minor third

Oscillator

This is a precision triangle oscillator with an external time constant (R x C). Capacitor $C_{\rm C}$ on pin $C_{\rm OSC}$ is charged by constant current to 1 V and then discharged to 0.5 V. The constant current is obtained on pin $R_{\rm OSC}$ with an external resistor $R_{\rm R}$ to ground.

When the voltage on $C_{\rm OSC}$ is building up, the logic is reset at 350 mV. This always ensures that a complete tone sequence is issued. If the oscillator pin is short-circuited to GND during operation, the sequence is repeated.

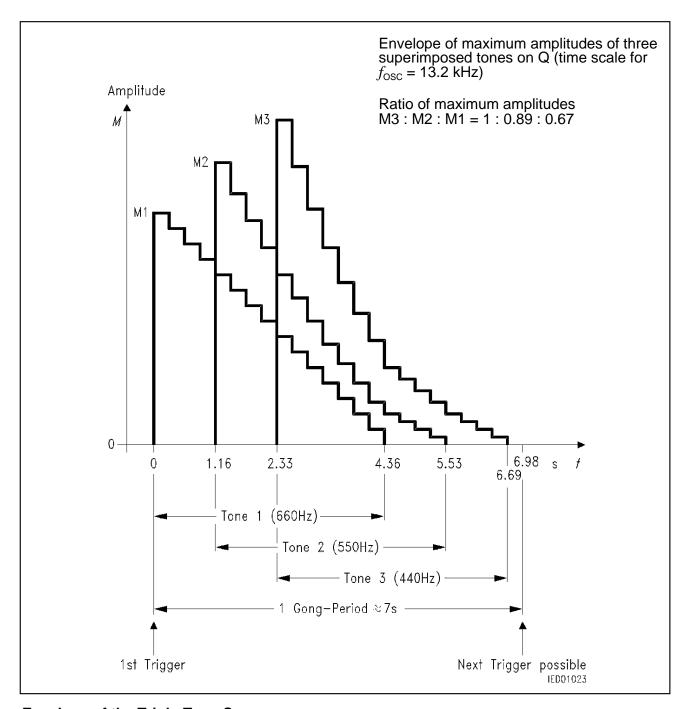


Voltages on Pin C_{osc}

Logic

The logic unit contains the complete sequence control. The oscillator produces the power-on reset and the clock frequency. Single-tone, dual-tone or triple-tone operation is programmed on inputs E1 and E2. The 4-bit digital/analog converters are driven in parallel. In the event of oscillator disturbance, and after the sequence, the dominant stop output is set. By applying current to pin L, the sequence can be shortened by a factor of 30 for test purposes.

The following figure shows the envelope of the triple-tone sequence:



Envelope of the Triple-Tone Sequence

Digital / Analog Converter, Loudness and Junction Control

The DAC converts the 4-bit words from the logic into the appropriate staircase currents with the particular tone frequency. The sum current $I_{\rm I}$ drives the following current amplifier. The loudness generator produces the DAC reference current $I_{\rm L}$ for all three tones. This requires connecting an external resistor to ground. The chip temperature is monitored by the junction control. At temperatures of more then approx. 170 °C the stop input will switch the output current $I_{\rm I}$ to zero. The output current is enabled again once the chip has cooled down to approx. 150 °C.

Current Amplifier

The current amplifier with a gain of 1600 boosts the current $I_{\rm I}$ from approx. 470 μ A maximum to approx. 750 mA maximum. The output stage consists of an NPN transistor with its emitter on power GND and collector on pin Q.

The current control insures that the output stage only conducts defined currents. In conjunction with the integrated thermal shutdown, this makes the configuration shortcircuit-protected within wide limits. Because of the absence of feedback the circuit is also extremely stable and therefore uncritical in applications. Resistor $R_{\rm L}$ on pin L sets the output voltage swing. This assumes that the resistive component of the loudspeaker impedance $R_{\rm Q}$ responds similarly as the resistance $R_{\rm L}$.

The output amplitude of the current $I_{\rm I}$ reaches the maximum $I_{\rm Imax} \cong 3$ x $V_{\rm L}$ / $R_{\rm L}$ at a time t of 2.33 s (only 3 tone mode), so $R_{\rm L}$ has to be scaled for this point.

The following applies:

$$I_{\rm Q} = I_{\rm Imax} \times {\rm B} = (V_{\rm S} - V_{\rm sat}) / R_{\rm Q} \approx 0.8 \ V_{\rm S} / R_{\rm Q}$$

3 x B x $(V_{\rm I} / R_{\rm I}) \approx 0.8 \ V_{\rm S} / R_{\rm Q}$

the result is:

$$R_{L} = R_{Q} \times 3 \times B \times (V_{L} / 0.8 V_{S})$$
 with: B = 1600
 $R_{I} = R_{Q} \times K \times (V_{I} / 0.8 V_{S})$ with: K = 4800



Application Hints and Application Circuit

1) Loudness Resistor (max. Load Current of 3-Tone Signal with Ensured Ratio of Amplitudes)

0.8
$$V_{\rm S}$$
 / $R_{\rm Q} \approx (V_{\rm L}$ / $R_{\rm L})$ x K

$$R_{L} = (V_{L} / 0.8 V_{S}) \times R_{Q} \times K; K = 4800$$

Example: $R_{\rm Q}$ = 8 Ω ; $V_{\rm S}$ = 5 V; $V_{\rm L}$ = 1.2 V

$$R_{L} = (1.2 / 4) \times 8 \Omega \times 4800 \approx 12 \text{ k}\Omega$$

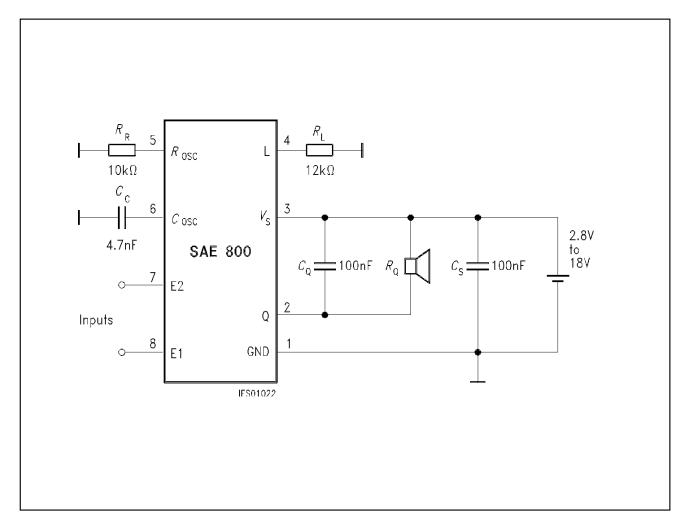
2) Oscillator Elements $R_{\rm R}$, $C_{ m C}$

$$f = 5 / 8 \times 1 / (R_R \times C_C)$$

Example: f = 13.2 kHz; $C_{\text{C}} = 4.7 \text{ nF}$

$$R_{\rm R}$$
 = 5 / (8 x 13.2 x 4.7) x 10 6 $\Omega \approx$ 10 $k\Omega$

The following is a typical application circuit



Application Circuit

Absolute Maximum Ratings

Parameter	Symbol	Limit Values		Unit
		min.	max.	
Supply voltage	V_{S}	- 0.3	24	V
Input voltage at E1, E2	$V_{E1,E2}$	-5	24	V
Current at output Q Current at input pins E1, E2	I_{Q} $I_{\mathrm{E1,E2}}$	- 50 - 2	750 3	mA mA
Current at pin $R_{\rm OSC}$ Current at pin L Current at pin $C_{\rm OSC}$	I_{R} I_{L} I_{C}	- 300 - 300 - 200	200 200 200	μΑ μΑ μΑ
Junction temperature	$T_{\rm j}$	- 50	150	°C
Storage temperature	$T_{ m stg}$	- 50	150	°C

Operating Range

Supply voltage	$V_{\mathtt{S}}$	2.8	18	V
Junction temperature	T_{j}	– 25	125	°C
Oscillator frequency at $C_{ m OSC}$	f_{C}		100	kHz
Current at pin R_{OSC} Current for test mode at pin L Current at pin L	$egin{array}{c} I_{R} \ I_{R} \ I_{L} \end{array}$	- 200 90 - 200	- 10 110 - 10	μΑ μΑ μΑ
Input voltage at E1, E2	$V_{E1,E2}$	- 4	18	V
Thermal resistance junction-air (P-DIP-8-4) junction-air (P-DSO-8-1)	$R_{thJA} \ R_{thJA}$		100 180	K/W K/W



Characteristics

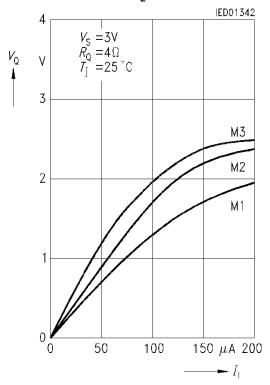
 $T_{\rm i}$ = -25 to 125°C; $V_{\rm S}$ = 2.8 to 18 V

Parameter	Symbol	Limit Values			Unit	Test
		min.	typ.	max.		Condition
Supply Section						
Standby current	I_{St}		1	10	μΑ	
Quiescent current; pin L open	I_{Qu}		5	10	mA	
Output Section						
Peak output power (tone 3)						
$V_{\rm S}$ = 2.8 V; $R_{\rm Q}$ = 4 Ω ; $R_{\rm L}$ = 8.2 k Ω	P_{Q}	250	330		mW	
$V_{\rm S}$ = 2.8 V; $R_{\rm Q}$ = 8 Ω ; $R_{\rm L}$ = 18 k Ω	P_{Q}	125	165		mW	
$V_{\rm S}$ = 5.0 V; $R_{\rm Q}$ = 8 Ω ; $R_{\rm L}$ = 10 k Ω	P_{Q}	450	600		mW	Α
$V_{\rm S}$ = 5.0 V; $R_{\rm Q}$ = 16 Ω ; $R_{\rm L}$ = 18 k Ω	P_{Q}	225	300		mW	
$V_{\rm S}$ = 12 V; $R_{\rm Q}$ = 50 Ω ; $R_{\rm L}$ = 33 k Ω	P_{Q}	450	600		mW	
Output level differences:						
tone 1 to 3	a_{13}	– 1		1	dB	A 1)
tone 2 to 3	a_{23}	- 1		1	dB	A ²⁾
Biasing Section Voltage at pin R_{OSC} ; R_{R} = 10 kΩ Voltage at pin L; R_{L} = 10 kΩ	$egin{array}{c} V_{R} \ V_{L} \end{array}$		1.2 1.2		V	
Oscillator Section			1		'	
Amplitude	$\Delta V_{ m C}$		0.5		V	
Frequency $R_{\rm R}$ = 10 k Ω ;	$ f_0 $		13.2		kHz	
$C_{\rm C} = 4.7 {\rm nF}$						
Oscill. drift vs. temperature	D_{T}	- 3		+ 3	10 ⁻⁴ /K	
Oscill. drift vs. supply voltage	D_{V}		1		10 ⁻³ /K	
Input Section						
Triggering voltage at E1, E2	V _{E1,E2}	1.6			V	
Triggering current at E1, E2	$I_{E1,E2}$	100			μA	
Noise voltage immunity at E1, E2	$V_{E1,E2}$			0.3	ľv	
Triggering delay at $f_0 = 13.2 \text{ kHz}$,	2	1	10	1	1

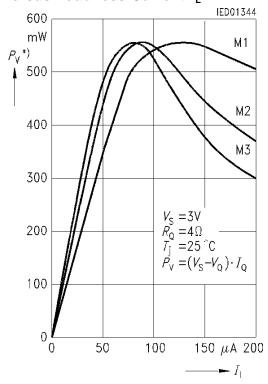
¹⁾ $a_{13} = 20 \text{ x log (M1 / (0.67 \text{ x M3)})}$

²⁾ $a_{23} = 20 \times \log (M2 / (0.89 \times M3))$

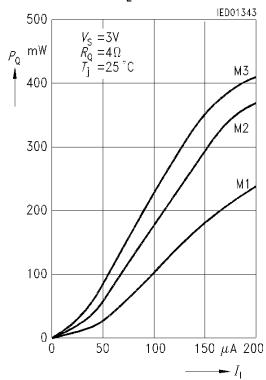
Output Peak Voltage $V_{\rm Q}$ versus Loudness-Current $I_{\rm L}$



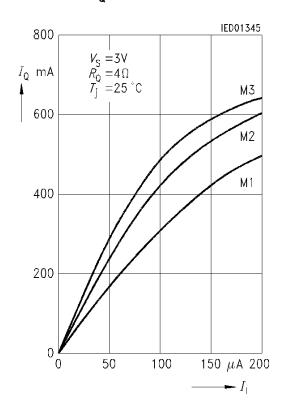
Power Dissipation $P_{\mathbf{v}}$ of Output Stage versus Loudness-Current $I_{\mathbf{L}}$



Max. Output Power P_{Q} versus Loudness-Current I_{L}

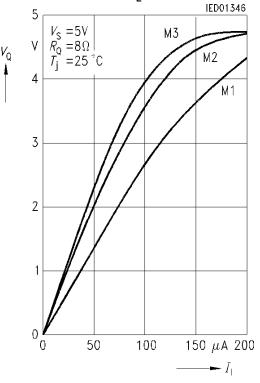


Peak Current I_Q versus Loudness-Current I_L



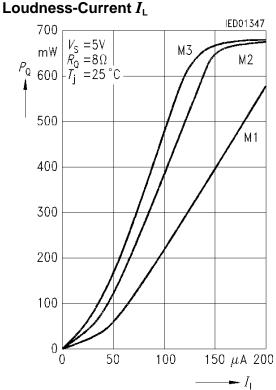
^{&#}x27;) Note that $I_{\rm Q}=f$ ($I_{\rm L}$) varies between 0 and K \cdot $I_{\rm L}$ during tone sequence. Thereby the maximum of the power dissipation during the tone sequence is the maximum of P $_{\rm v}$ (in diagram) between $I_{\rm L}=0$ and chosen $I_{\rm L}=V_{\rm L}/R_{\rm L}$.

Output Peak Voltage V_{Q} versus Loudness-Current I_{L}

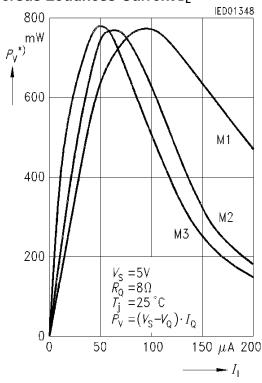


700 М3

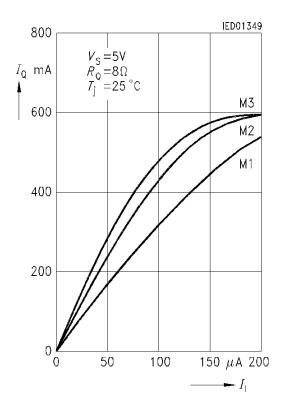
Max. Output Power P_{Q} versus



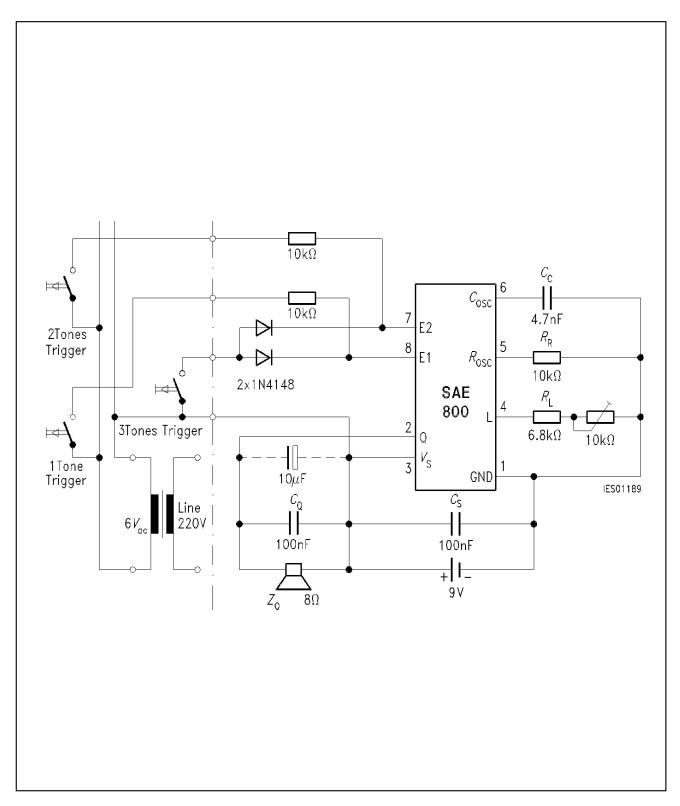
Power Dissipation P_v of Output Stage versus Loudness-Current $I_{\scriptscriptstyle m L}$



Peak Current I_Q versus Loudness-Current I_L

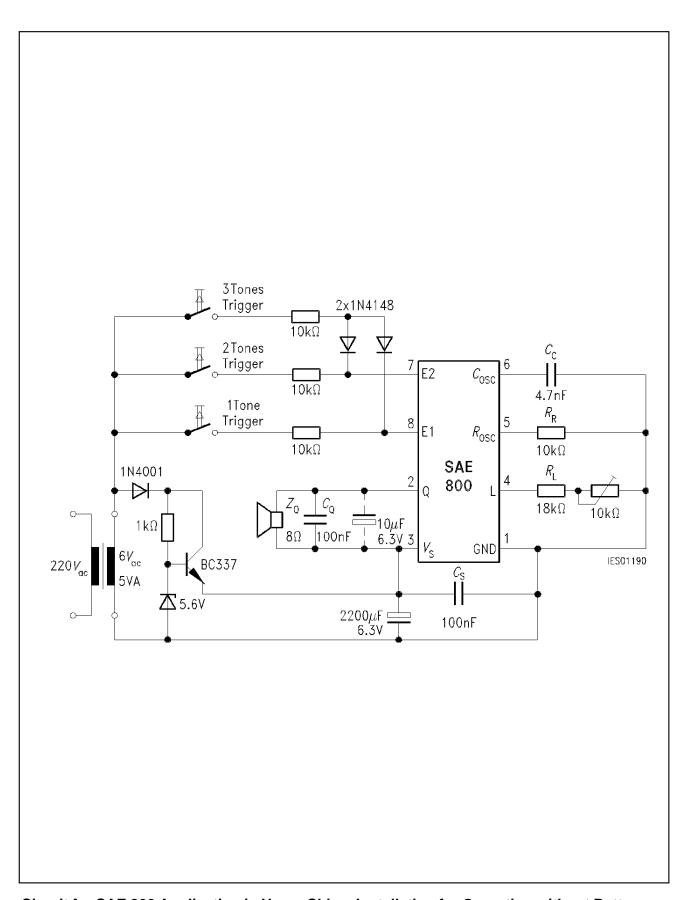


^{*)} Note that $I_{\rm Q}=f\left(I_{\rm L}\right)$ varies between 0 and K \cdot $I_{\rm L}$ during tone sequence. Thereby the maximum of the power dissipation during the tone sequence is the maximum of P_v (in diagram) between $I_{\rm L}$ = 0 and chosen $I_{\rm L}=V_{\rm L}/R_{\rm L}$.



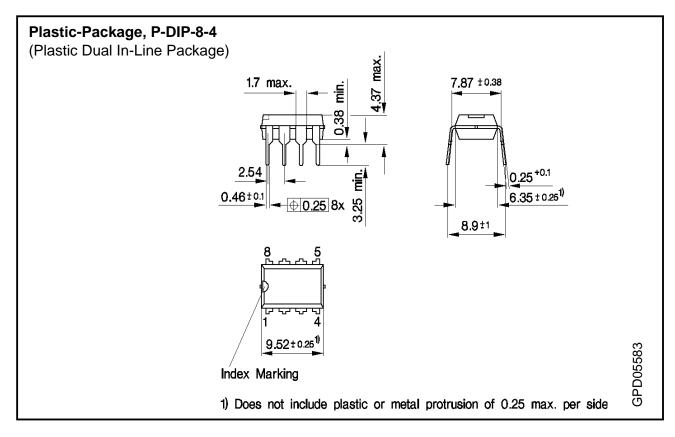
Circuit for SAE 800 Application in Home Chime Installation Utilizing AC and DC Triggering for 1, 2 or 3 Tone Chime; Adjustable Volume

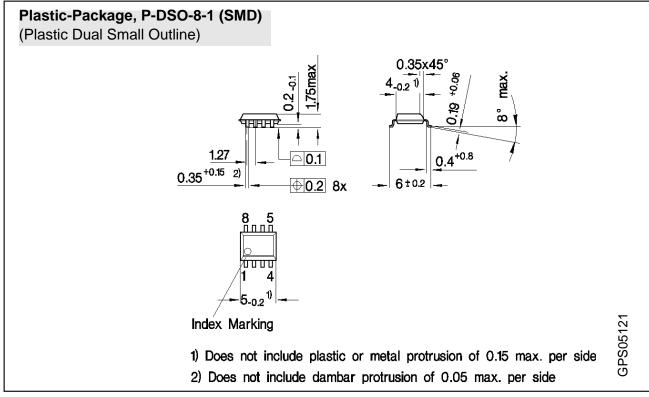
PCB layout information: Because of the peak currents at $V_{\rm S}$, Q and GND the lines should be designed in a flatspread way or as star pattern.



Circuit for SAE 800 Application in Home Chime Installation for Operation without Battery

Package Outlines





SMD = Surface Mounted Device

Dimensions in mm