

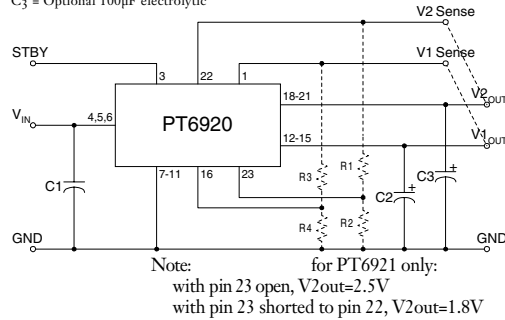
Patent Pending*

- Dual Outputs:
+3.3V/6A
+2.5V/2.2A or +1.8V/1.5A
- Adjustable Output Voltage
- Remote Sense (both outputs)
- Standby Function
- Over-Temperature Protection
- Soft-Start
- Internal Sequencing
- 23-pin SIPP Package

The PT6920 is a series of 25W dual output ISRs that were purposely designed to power the latest generation DSP chips. Both output voltages are independently adjustable, allowing either output voltages to be changed to accommodate a DSP upgrade. The internal power sequencing of both outputs meet the requirements of TI's 'C6000 series DSPs.

Standard Application

C₁ = Req'd 560µF electrolytic (1)
C₂ = Req'd 330µF electrolytic (1)
C₃ = Optional 100µF electrolytic



Pin-Out Information

Pin	Function	Pin	Function
1	V ₁ Remote Sense	13	V _{1out}
2	Do Not Connect	14	V _{1out}
3	STBY	15	V _{1out}
4	V _{in}	16	V ₁ Adjust
5	V _{in}	17	Do Not Connect
6	V _{in}	18	V _{2out}
7	GND	19	V _{2out}
8	GND	20	V _{2out}
9	GND	21	V _{2out}
10	GND	22	V ₂ Remote Sense
11	GND	23	V ₂ Adjust*
12	V _{1out}		

Ordering Information

PT6921□ = +3.3 Volts
+2.5/+1.8 Volts
PT6922□ = +3.3 Volts
+1.5 Volts

PT Series Suffix (PT1234X)

Case/Pin
Configuration

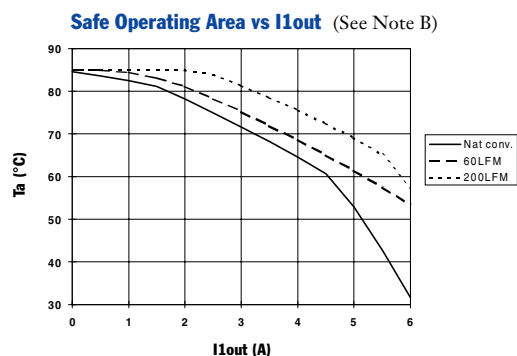
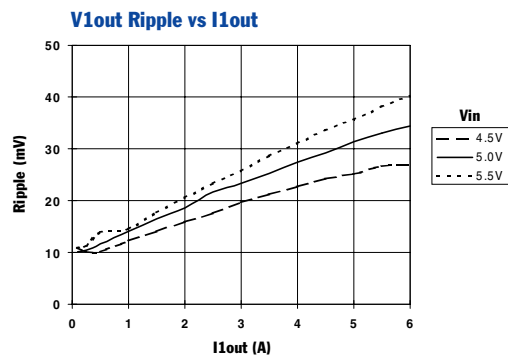
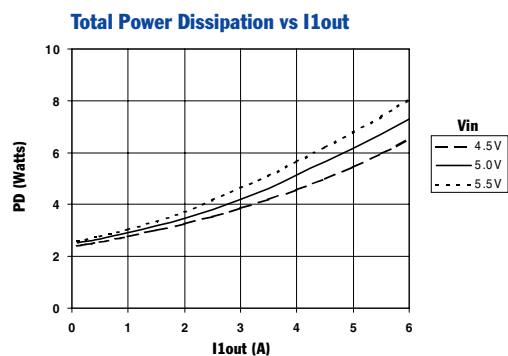
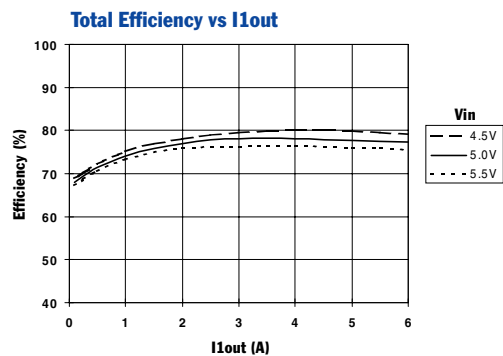
Vertical Through-Hole	N
Horizontal Through-Hole	A
Horizontal Surface Mount	C

(For dimensions and PC board layout, see Package Styles 1100 and 1110.)

Specifications

Characteristics (T _a = 25°C unless noted)	Symbols	Conditions	PT6920 SERIES			Units	
			Min	Typ	Max		
Output Current	I _o	T _a = +60°C, 200 LFM, pkg N	V ₁ = 3.3V	0.1 ⁽²⁾	—	5.5 ⁽³⁾	A
			V ₂ = 2.5V	0	—	2.2 ⁽³⁾	
			V ₂ = 1.8V	0	—	1.75 ⁽³⁾	
			V ₂ = 1.2V	0	—	1.2 ⁽³⁾	
		T _a = +25°C, natural convection	V ₁ = 3.3V	0.1	—	6.0	A
			V ₂ = 2.5V	0	—	2.2	
V ₂ = 1.8V	0		—	1.75			
V ₂ = 1.2V	0		—	1.2			
Input Voltage Range	V _{in}	0.1A ≤ I _o ≤ I _{max}	4.5	—	5.5	V	
Output Voltage Tolerance	ΔV _o	V _{in} = +5V, I _o = I _{max} , both outputs 0°C ≤ T _a ≤ +65°C	V _o -0.1	—	V _o +0.1	V	
Line Regulation	Reg _{line}	4.5V ≤ V _{in} ≤ 5.5V, I _o = I _{max}	V ₁ = 3.3V V ₂ = 2.5V	— —	±7 ±7	±17 ±13	mV
Load Regulation	Reg _{load}	V _{in} = +5V, 0.1 ≤ I _o ≤ I _{max}	V ₁ = 3.3V V ₂ = 2.5V	— —	±17 ±4	±33 ±10	mV
V _o Ripple/Noise	V _n	V _{in} = +5V, I _o = I _{max}	V ₁ = 3.3V V ₂ = 2.5V	— —	50 25	— —	mV
Transient Response with C ₂ = 330pF	t _{tr} V _{os}	I _o step between 0.5I _{Lmax} and I _{Lmax} V _o over/undershoot	V ₁ = 3.3V	—	25	—	μSec
			V ₂ = 2.5V	—	60	—	mV
				—	60	—	
Efficiency	η	V _{in} = +5V, I _o = 4A total	—	75	—	%	
Switching Frequency	f _o	4.5V ≤ V _{in} ≤ 5.5V 0.1A ≤ I _o ≤ I _{max}	475	600	725	kHz	
Absolute Maximum Operating Temperature Range	T _a	Over V _{in} Range	−40 ⁽⁴⁾	—	+85 ⁽⁵⁾	°C	
Storage Temperature	T _s	—	−40	—	+125	°C	
Weight	—	Vertical/Horizontal	—	29	—	grams	

- Notes:** (1) The PT6920 series requires a 560µF electrolytic capacitor on the input and a 330µF electrolytic capacitor on the output for proper operation in all applications.
(2) I_{min} current of 0.1A can be divided between both outputs; V₁, or V₂. The ISR will operate down to no-load with reduced specifications.
(3) I_{max} listed for each output assumes the maximum current drawn simultaneously on both outputs. Consult the factory for the absolute maximum.
(4) For operating temperatures below 0°C, use tantalum type capacitors on both the input and output.
(5) See Safe Operating Area curves for appropriate derating.

PT6921, $V_{2out} = 2.5V$, $I_{2out} = 2.2A$ (See Note A)

Note A: All characteristic data listed in the above graphs has been developed from actual products tested at 25°C. This data is considered typical data for the ISR.

Note B: SOA curves represent operating conditions at which internal components are at or below manufacturer's maximum rated operating conditions.

Adjusting the Output Voltage of the PT6920 and PT6930 Dual Output Voltage ISRs

Each output voltage from the PT6920 and PT6930 series of ISRs can be independently adjusted higher or lower than the factory trimmed pre-set voltage. V_1 (the voltage at V_1 out), or V_2 (the voltage at V_2 out) may each be adjusted either up or down using a single external resistor ². Table 1 gives the adjustment range for both V_1 and V_2 for each model in the series as V_a (min) and V_a (max). Note that V_2 must always be lower than V_1 ³.

V_1 Adjust Up: To increase the output, add a resistor R4 between pin 16 (V_1 Adjust) and pins 7-11 (GND) ².

V_1 Adjust Down: Add a resistor (R3), between pin 16 (V_1 Adjust) and pin 1 (V_1 Remote Sense) ².

V_2 Adjust Up: Add a resistor R2 between pin 23 (V_2 Adjust) and pins 7-11 (GND) ².

V_2 Adjust Down: Add a resistor (R1) between pin 23 (V_2 Adjust) and pin 22 (V_2 Remote Sense) ².

Refer to Figure 1 and Table 2 for both the placement and value of the required resistor.

Notes:

1. The output voltages, V_1 out and V_2 out, may be adjusted independently.
2. Use only a single 1% resistor in either the (R3) or R4 location to adjust V_1 , and in the (R1) or R2 location to adjust V_2 . Place the resistor as close to the ISR as possible.
3. V_2 must always be at least 0.2V lower than V_1 .
4. V_2 on both the PT6921 and PT6931 models may be adjusted from 2.5V to 1.8V by simply connecting pin 22 (V_2 Remote Sense) to pin 23 (V_2 Adjust). For more details, consult the data sheet.

5. If V_1 is increased above 3.3V, the minimum input voltage to the ISR must also be increased. The minimum required input voltage must be $(V_1 + 1.2)V$ or 4.5V, whichever is greater. Do not exceed 5.5V
6. Never connect capacitors to either the V_1 Adjust or V_2 Adjust pins. Any capacitance added to these control pins will affect the stability of the respective regulated output.
7. Adjusting either voltage (V_1 or V_2) may increase the power dissipation in the regulator, and correspondingly change the maximum current available at either output. Consult the factory for application assistance.

The adjust up and adjust down resistor values can also be calculated using the following formulas. Be sure to select the correct formula parameter from Table 1 for the output and model being adjusted.

$$(R1) \text{ or } (R3) = \frac{R_o(V_a - 1)}{V_o - V_a} - R_s \quad \text{k}\Omega$$

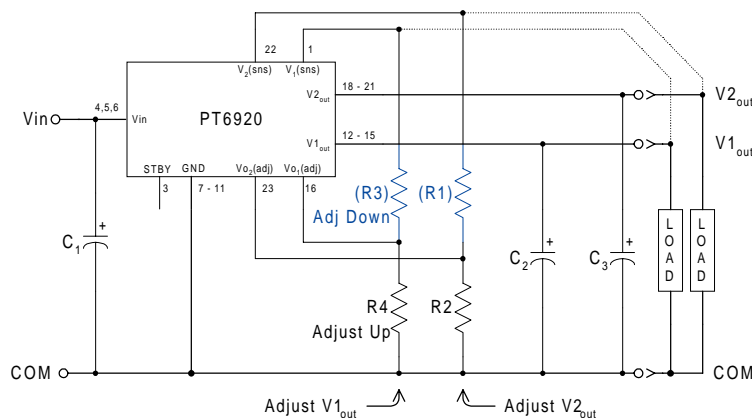
$$R2 \text{ or } R4 = \frac{R_o}{V_a - V_o} - R_s \quad \text{k}\Omega$$

Where: V_o = Original output voltage, (V_1 or V_2)
 V_a = Adjusted output voltage
 R_o = The resistance value from Table 1
 R_s = The series resistance from Table 1

Table 1

PT6920 ADJUSTMENT RANGE AND FORMULA PARAMETERS			
Output Bus	V_1 out	V_2 out	
Series Pt #			
Standard Case	PT6921/22	PT6921	PT6922
Excalibur Case	PT6931/32	PT6931	PT6932
Adj. Resistor	(R3)/R4	(R1)/R2	(R1)/R2
V_o (nom)	3.3V	2.5V	1.5
V_a (min)	2.3V	1.8V	1.2
V_a (max)	3.6V	3.0V	3.0
R_o (k Ω)	12.1	10.0	9.76
R_s (k Ω)	12.1	11.5	6.49

Figure 1



PT6920/PT6930 Series

Table 2

PT6920/PT6930 ADJUSTMENT RESISTOR VALUES

Output Bus	V _{1out}	V _{2out}	
Series Pt#			
Standard Case	PT6921/6922	PT6921	PT6922
Excalibur Case	PT6931/6932	PT6931	PT6932
Adj Resistor	(R3)/R4	(R1)/R2	(R1)/R2
V ₀ (nom)	3.3Vdc	2.5Vdc	1.5Vdc
V _A (req'd)			
1.2			(0.0)kΩ
1.25			(3.3)kΩ
1.3			(8.2)kΩ
1.35			(16.3)kΩ
1.4			(32.6)kΩ
1.45			(81.4)kΩ
1.5			
1.55			189.0kΩ
1.6			91.1kΩ
1.65			58.6kΩ
1.7			42.3kΩ
1.75			32.6kΩ
1.8		(0.0)kΩ	26.0kΩ
1.85		(1.6)kΩ	21.4kΩ
1.9		(3.5)kΩ	17.9kΩ
1.95		(5.8)kΩ	15.2kΩ
2.0		(8.5)kΩ	13.0kΩ
2.05		(11.8)kΩ	11.3kΩ
2.1		(16.0)kΩ	9.8kΩ
2.15		(21.4)kΩ	8.5kΩ
2.2		(28.5)kΩ	7.5kΩ
2.25		(38.5)kΩ	6.5kΩ
2.3	(3.6)kΩ	(53.5)kΩ	5.7kΩ
2.35	(5.1)kΩ	(78.5)kΩ	5.0kΩ
2.4	(6.7)kΩ	(129.0)kΩ	4.4kΩ
2.45	(8.5)kΩ	(279.0)kΩ	3.8kΩ
2.5	(10.6)kΩ		3.3kΩ
2.55	(12.9)kΩ	189.0kΩ	2.8kΩ
2.6	(15.6)kΩ	88.5kΩ	2.4kΩ
2.65	(18.6)kΩ	55.2kΩ	2.0kΩ
2.7	(22.2)kΩ See Note 3	38.5kΩ	1.6kΩ
2.75	(26.4)kΩ	28.5kΩ	1.3kΩ
2.8	(31.5)kΩ	21.8kΩ	1.0kΩ
2.85	(37.6)kΩ	17.1kΩ	0.7kΩ
2.9	(45.4)kΩ	13.5kΩ	0.5kΩ
2.95	(55.3)kΩ	10.7kΩ	0.2kΩ
3.0	(68.6)kΩ	8.5kΩ	0.0kΩ
3.05	(87.1)kΩ		
3.1	(115.0)kΩ		
3.15	(161.0)kΩ		
3.2	(254.0)kΩ		
3.25	(532.0)kΩ		
3.3			
3.4	109.0kΩ See Note 5		
3.5	48.4kΩ		
3.6	28.2kΩ		

R1/R3 = (Blue) R2/R4 = Black

Using the Standby Function on the PT6920 and PT6930 Dual Output Voltage Converters

Both output voltages of the 23-pin PT6920/6930 dual output converter may be disabled using the regulator's standby function. This function may be used in applications that require power-up/shutdown sequencing, or wherever there is a requirement to control the output voltage On/Off status with external circuitry.

The standby function is provided by the *STBY** control, pin 3. If pin 3 is left open-circuit the regulator operates normally, and provides a regulated output at both V_{1out} (pins 12–15) and V_{2out} (pins 18–21) whenever a valid supply voltage is applied to V_{in} (pins 4, 5, & 6) with respect to GND (pins 7–11). If a low voltage² is then applied to pin-3 both regulator outputs will be simultaneously disabled and the input current drawn by the ISR will typically drop to less than 30mA (50mA max). The standby control may also be used to hold-off both regulator outputs during the period that input power is applied.

The standby pin is ideally controlled using an open-collector (or open-drain) discrete transistor (See Figure 1). It may also be driven directly from a dedicated TTL³ compatible gate. Table 1 provides details of the threshold requirements.

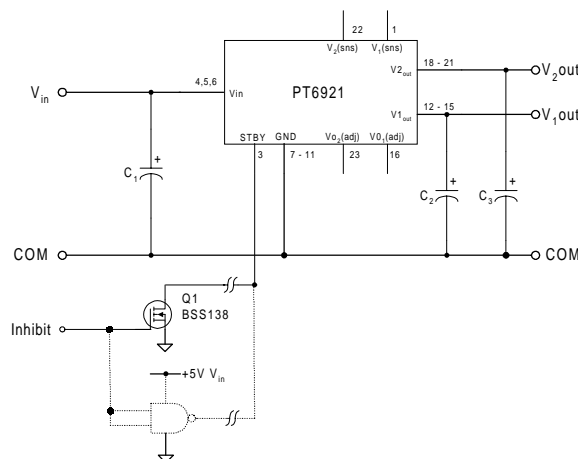
Table 1 Inhibit Control Thresholds^{2,3}

Parameter	Min	Max
Enable (V_{IH})	1.8V	V_{in}
Disable (V_{IL})	-0.1V	0.8V

Notes:

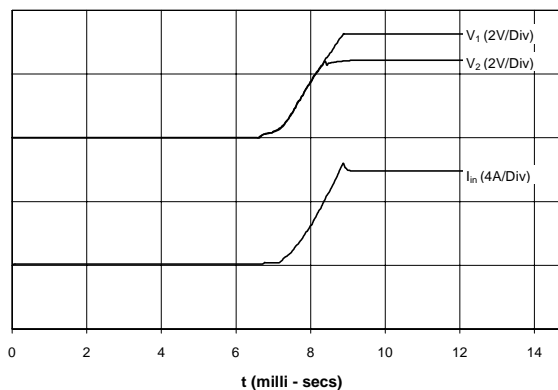
1. The Standby/Inhibit control logic is similar for all Power Trends' modules, but the flexibility and threshold tolerances will be different. For specific information on this function for other regulator models, consult the applicable application note.
2. The Standby control pin is ideally controlled using an open-collector (or open-drain) discrete transistor and requires no external pull-up resistor. To disable the regulator output, the control pin must be pulled to less than 0.8Vdc with a low-level 0.5mA sink to ground.
3. The Standby input on the PT6920/6930 series may be driven by a differential output device, making it directly compatible with TTL logic. The control input has an internal pull-up to the input voltage V_{in} . A voltage of 1.8V or greater ensures that the regulator is enabled. Do not use devices that can drive the Standby control input above 5.5V or V_{in} .

Figure 1



Turn-On Time: Turning Q_1 in Figure 1 off removes the low-voltage signal at pin 3 and enables both outputs from the PT6920/6930 regulator. Following a delay of about 5–10ms, V_{1out} and V_{2out} rise together until the lower voltage, V_{2out} , reaches its set output. V_{1out} then continues to rise until both outputs reach full regulation voltage. The total power-up time is less than 15ms, and is relatively independent of load, temperature, and output capacitance. Figure 2 shows waveforms of the input current I_{in} , and output voltages V_{1out} and V_{2out} , for a PT6921 (3.3V/2.5V). The turn-off of Q_1 corresponds to $t = 0$ secs. The waveforms were measured with a 5Vdc input voltage, and with resistive loads of 5.5A and 2.2A at the V_{1out} and V_{2out} outputs respectively.

Figure 2



PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Samples (Requires Login)
PT6921A	LIFEBUY	SIP MODULE	EJJ	23	8	Pb-Free (RoHS)	Call TI	N / A for Pkg Type	
PT6921C	LIFEBUY	SIP MODULE	EJK	23	8	Pb-Free (RoHS)	Call TI	Level-1-215C-UNLIM	
PT6921N	LIFEBUY	SIP MODULE	EJH	23	10	Pb-Free (RoHS)	Call TI	N / A for Pkg Type	

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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