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February 2016

# FSL137H Green Mode Fairchild Power Switch (FPS™)

### **Features**

- Built-in 5ms Soft-Start Function
- Internal Avalanche Rugged 700V SenseFET
- Low Audio Noise
- High-Voltage Startup
- Fixed PWM Frequency at 100KHz
- Linearly Decreasing PWM Frequency to 18KHz
- Peak-Current-Mode Control
- Cycle-by-Cycle Current Limiting
- Leading-Edge Blanking (LEB)
- Synchronized Slope Compensation
- Internal Open-loop Protection (OLP)
- V<sub>DD</sub> Under-Voltage Lockout (UVLO)
- V<sub>DD</sub> Over-Voltage Protection (OVP)
- Constant Power Limit (Full AC Input Range)
- Internal OTP Sensor with Hysteresis

### **Applications**

General-purpose switch-mode power supplies and flyback power converters, including:

- SMPS for VCR, SVR, STB, DVD & VCD Player, Printer, Facsimile, & Scanner
- Adapter for Camcorder

### **Description**

The highly integrated FSL137H consists of an integrated current mode Pulse Width Modulator (PWM) and an avalanche-rugged 700V SenseFET. It is specifically designed for high-performance offline Switch Mode Power Supplies (SMPS) with minimal external components.

The integrated PWM controller features include a proprietary green-mode function that provides off-time modulation to linearly decrease the switching frequency at light-load conditions to minimize standby power consumption. To avoid acoustic noise problems, the minimum PWM frequency is set above 18KHz. The green-mode function enables the power supply to meet international power conservation requirements. With the internal high-voltage startup circuitry, the power loss due to bleeding resistors is also eliminated. To further reduce power consumption, the PWM controller is manufactured using the BiCMOS process, which allows an operating current of only 3.5mA.

The FSL137H built-in synchronized slope compensation achieves stable peak-current-mode control. The proprietary external line compensation ensures constant output power limit over a wide AC input voltage range, from  $90V_{AC}$  to  $264V_{AC}.$ 

The FSL137H provides many protection functions. In addition to cycle-by-cycle current limiting, the internal open-loop protection circuit ensures safety when an open-loop or output short-circuit failure occurs. PWM output is disabled until  $V_{DD}$  drops below the UVLO lower limit, when the controller starts up again. As long as  $V_{DD}$  exceeds ~28V, the internal OVP circuit is triggered.

Compared to a discrete MOSFET and controller or RCC switching converter solution, the FSL137H reduces total component count, design size, and weight while increasing efficiency, productivity, and system reliability. These devices provide a basic platform well suited for design of cost-effective flyback converters.

### **Ordering Information**

Part Number	SenseFET	Operating Temperature Range	Package	Packing Method
FSL137HNY	3.0A 700V	-40°C to +105°C	8-Pin Dual In-Line Package (DIP)	Tube

### **Application Diagram**

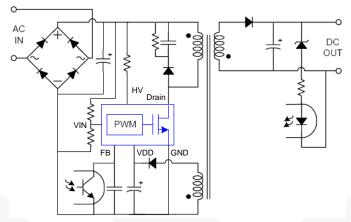


Figure 1. Typical Flyback Application

## Output Power Table<sup>(1)</sup>

Product	230V <sub>AC</sub>	± 15% <sup>(2)</sup>	85-265V <sub>AC</sub>		
Product	Adapter <sup>(3)</sup>	Open Frame <sup>(4)</sup>	Adapter <sup>(3)</sup>	Open Frame <sup>(4)</sup>	
FSL137H	17.5W	25W	13W	19W	

#### Notes:

- 1. The maximum output power can be limited by junction temperature.
- 2. 230  $V_{AC}$  or 100/115  $V_{AC}$  with doublers.
- Typical continuous power in a non-ventilated enclosed adapter with sufficient drain pattern as a heat sink, at T<sub>A</sub>=50°C ambient.
- 4. Maximum practical continuous power in an open-frame design with sufficient drain pattern as a heat sink, at  $T_A$ =50°C ambient.

### **Internal Block Diagram**

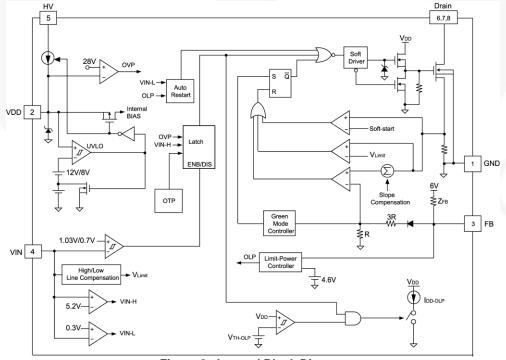
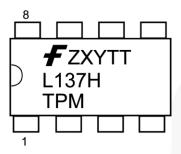


Figure 2. Internal Block Diagram

### **Pin Configuration**



F - Fairchild Logo

Z - Plant Code

X – 1-Digit Year Code Y – 1-Digit Week Code

TT – 2-Digit Die Run Code

T – Package Type (N: DIP)
P – Y: Green Package
M – Manufacture Flow Code

Figure 3. Pin Configuration

### **Pin Definitions**

Pin #	Name	Description				
1	GND	iround. SenseFET source terminal on primary side and internal controller ground.				
2	VDD	<b>Power Supply</b> . The internal protection circuit disables PWM output as long as $V_{DD}$ exceeds the OVP trigger point.				
3	FB	<b>Feedback</b> . The signal from the external compensation circuit is fed into this pin. The PWM durcycle is determined in response to the signal on this pin and the internal current-sense signal.				
4	VIN	<b>Line-Voltage Detection</b> . The line-voltage detection is used for brownout protection with hysteresis and constant output power limit over universal AC input range. This pin has additional protections that are pull-HIGH latch and pull-low auto recovery, depending on the application.				
5	HV	Startup. For startup, this pin is pulled HIGH to the line input or bulk capacitor via resistors.				
6	Drain	SenseFET Drain. High-voltage power SenseFET drain connection.				
7	Drain	SenseFET Drain. High-voltage power SenseFET drain connection.				
8	Drain	SenseFET Drain. High-voltage power SenseFET drain connection.				

### **Absolute Maximum Ratings**

Stresses exceeding the absolute maximum ratings may damage the device. The device may not function or be operable above the recommended operating conditions and stressing the parts to these levels is not recommended. In addition, extended exposure to stresses above the recommended operating conditions may affect device reliability. The absolute maximum ratings are stress ratings only.

Symbol	Par	ameter	Min.	Max.	Unit
V <sub>DRAIN</sub>	Drain Pin Voltage <sup>(5, 6)</sup>		700	V	
I <sub>DM</sub>	Drain Current Pulsed <sup>(7)</sup>			12	Α
E <sub>AS</sub>	Single Pulsed Avalanche Energy <sup>(8)</sup>			230	mJ
$V_{VDD}$	DC Supply Voltage			30	V
V <sub>FB</sub>	FB Pin Input Voltage		-0.3	7.0	V
$V_{VIN}$	VIN Pin Input Voltage		-0.3	7.0	V
$V_{HV}$	HV Pin Input Voltage			700	V
P <sub>D</sub>	Power Dissipation (T <sub>A</sub> <50°C)	-/-		1.5	W
$\theta_{JA}$	Junction-to-Air Thermal Resistance			80	°C/W
Ψл	Junction-to-Top Thermal Resistance	e <sup>(9)</sup>		35	°C/W
TJ	Operating Junction Temperature			+150	°C
T <sub>STG</sub>	Storage Temperature Range		-55	150	°C
TL	Lead Temperature (Wave Soldering	g or IR, 10 Seconds)	No.	+260	°C
ESD	Electrostatic Discharge Capability, All Pins Except HV Pin <sup>(10)</sup> Human Body Model: JESD22-A114 Charged Device Model: JESD22-C101		Λ,,	4.5	kV
ESD				1.5	ΚV

#### Notes:

- 5. All voltage values, except differential voltages, are given with respect to the network ground terminal.
- 6. Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device.
- 7. Non-repetitive rating: Pulse width is limited by maximum junction temperature.
- 8. L = 51mH, starting  $T_J = 25$ °C.
- 9. Measured on the package top surface.
- 10. All pins including HV pin: HBM=1kV, CDM=1.25kV

### **Recommended Operating Conditions**

The Recommended Operating Conditions table defines the conditions for actual device operation. Recommended operating conditions are specified to ensure optimal performance to the datasheet specifications. Fairchild does not recommend exceeding them or designing to Absolute Maximum Ratings.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
T <sub>A</sub>	Operating Ambient Temperature		-40		+105	°C

### **Electrical Characteristics**

 $V_{DD}{=}15V,\,T_{A}{=}25^{\circ}C$  unless otherwise specified.

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
SenseFET S	Section <sup>(11)</sup>	1				
$BV_{DSS}$	Drain-Source Breakdown Voltage	V <sub>GS</sub> = 0V	700			V
		$V_{DS} = 700V, V_{GS} = 0V$		0.5	50.0	
I <sub>DSS</sub>	Zero-Gate-Voltage Drain Current	$V_{DS} = 560V, V_{GS} = 0V,$ $T_A = 125$ °C		1	200	μA
R <sub>DS(ON)</sub>	Drain-Source On-State Resistance <sup>(12)</sup>	$V_{GS} = 10V, I_D = 0.5A$		4.00	4.75	Ω
C <sub>ISS</sub>	Input Capacitance	$V_{GS} = 0V$ , $V_{DS} = 25V$ , $f = 1MHz$		315	410	pF
Coss	Output Capacitance	$V_{GS} = 0V$ , $V_{DS} = 25V$ , $f = 1MHz$	1	47	61	pF
C <sub>RSS</sub>	Reverse Transfer Capacitance	$V_{GS} = 0V$ , $V_{DS} = 25V$ , $f = 1MHz$		9	14	pF
$t_{d(on)}$	Turn-On Delay Time	$V_{DS} = 350V, I_{D} = 1.0A$		11.2	33.0	ns
t <sub>r</sub>	Rise Time	$V_{DS} = 350V, I_{D} = 1.0A$		34	78	ns
$t_{\text{d(off)}}$	Turn-Off Delay Time	$V_{DS} = 350V, I_{D} = 1.0A$	)	28.2	67.0	ns
t <sub>f</sub>	Fall Time	$V_{DS} = 350V, I_{D} = 1.0A$		32	74	ns
V <sub>DD</sub> Section	i y					
V <sub>OP</sub>	Continuously Operating Voltage				22	V
$V_{\text{DD-ON}}$	Start Threshold Voltage		11	12	13	V
$V_{DD-OFF}$	Minimum Operating Voltage		7	8	9	V
I <sub>DD-ST</sub>	Startup Current	V <sub>DD-ON</sub> – 0.16V			30	μΑ
I <sub>DD-OP</sub>	Operating Supply Current	$V_{DD} = 15V, V_{FB} = 3V$	3.0	3.5	4.0	mA
I <sub>DD-BM</sub>	Green-Mode Operating Supply Current	$V_{FB} = V_{FB-G}$		2		mA
I <sub>DD-OLP</sub>	Internal Sink Current	V <sub>TH-OLP</sub> +0.1V	30	60	90	μA
V <sub>TH-OLP</sub>	I <sub>DD-OLP</sub> Off Voltage		5	6	7	V
V <sub>DD-OVP</sub>	V <sub>DD</sub> Over-Voltage Protection		27	28	29	V
t <sub>D-VDDOVP</sub>	V <sub>DD</sub> Over-Voltage Protection Debounce Time		75	130	200	μs
<b>HV Section</b>					- 1	
I <sub>HV</sub>	Maximum Current Drawn from HV Pin	HV 120V <sub>DC</sub> , $V_{DD} = 0V$ with $10\mu F$	1.5	3.5	5.0	mA
I <sub>HV-LC</sub>	Leakage Current After Startup	$HV = 700V,$ $V_{DD} = V_{DD-OFF} + 1V$		1	20	μА
Oscillator S	Section					
fosc	Frequency in Nominal Mode	Center Frequency	94	100	106	kHz
f <sub>OSC-G</sub>	Green-Mode Frequency		14	18	22	kHz
D <sub>MAX</sub>	Maximum Duty Cycle			85		%
f <sub>DV</sub>	Frequency Variation vs. V <sub>DD</sub> Deviation	V <sub>DD</sub> = 9V to 22V			5	%
f <sub>DT</sub>	Frequency Variation vs. Temperature Deviation <sup>(11)</sup>	$T_A = -40 \text{ to } +105^{\circ}\text{C}$			5	%

Continued on the following page...

### **Electrical Characteristics** (Continued)

 $V_{DD}$ =15V,  $T_A$ =25°C unless otherwise specified.

Symbol	Parameter Condition		Min.	Тур.	Max.	Unit
V <sub>IN</sub> Section						I
V <sub>IN-ON</sub>	PWM Turn-on Threshold Voltage		0.98	1.03	1.08	V
V <sub>IN-RL</sub>	Release Latch Voltage		0.65	0.70	0.75	V
V <sub>IN-H</sub>	Pull HIGH Latch Trigger Level		4.9	5.2	5.5	V
t <sub>IN-H</sub>	Pull HIGH Latch Debounce Time			100		μs
$V_{IN-L}$	Pull LOW Auto Recovery Trigger Level		0.2	0.3	0.4	V
Feedback Ir	nput Section					I
Av	FB Voltage to Current-Sense Attenuation			1/4.0		V/V
$Z_{FB}$	Input Impedance			9.5		kΩ
V <sub>FB-OPEN</sub>	Output High Voltage		5			V
$V_{FB-OLP}$	FB Open-Loop Trigger Level		4.4	4.6	4.8	V
t <sub>D-OLP</sub>	Delay Time of FB Pin Open-loop Protection		50	56	59	ms
$V_{FB-N}$	Green-Mode Entry FB Voltage		2.3	2.5	2.7	V
$V_{FB-G}$	Green-Mode Ending FB Voltage			V <sub>FB-N</sub> - 0.1		V
$V_{FB-ZDC}$	Zero Duty Cycle FB Voltage		1.9	2.1	2.3	V

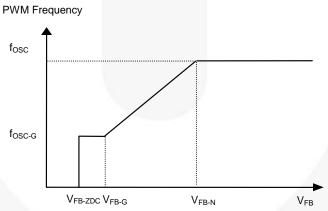


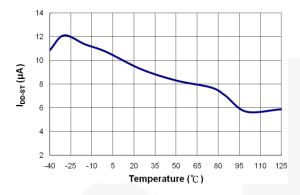
Figure 4. V<sub>FB</sub> vs. PWM Frequency

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit	
Current-Sense Section							
I <sub>LIM</sub> at V <sub>IN</sub> = 1.2V	Peak Current Limit	V <sub>IN</sub> = 1.2V	0.74	0.84	0.94	Α	
$I_{LIM}$ at $V_{IN} = 3.6V$	Peak Current Limit	V <sub>IN</sub> = 3.6V	0.64	0.74	0.84	Α	
t <sub>SS</sub>	Period during Soft Startup Time <sup>(11)</sup>	Period during Soft Startup Time <sup>(11)</sup>		5.0	5.5	ms	
Over-Temperature Protection Section (OTP)							
T <sub>OTP</sub>	Protection Junction Temperature <sup>(11,13)</sup>			142		°C	

#### Notes:

- 11. These parameters, although guaranteed, are not 100% tested in production.
- 12. Pulse test: pulse width  $\leq 300 \mu s$ , duty  $\leq 2\%$ .
- 13. When activated, the output is disabled and the latch is turned off.

### **Typical Characteristics**



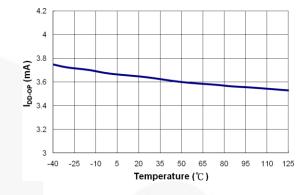
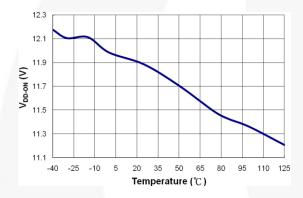


Figure 5. I<sub>DD-ST</sub> vs. Temperature





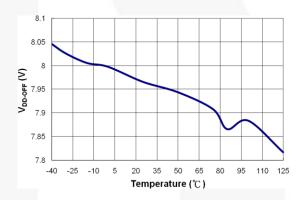
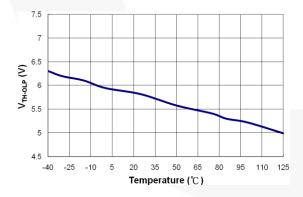


Figure 7. V<sub>DD-ON</sub> vs. Temperature

Figure 8. V<sub>DD-OFF</sub> vs. Temperature



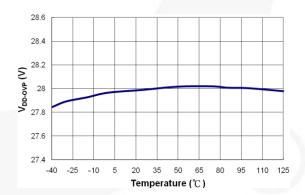


Figure 9. V<sub>TH-OLP</sub> vs. Temperature

Figure 10. V<sub>DD-OVP</sub> vs. Temperature

### **Typical Characteristics**

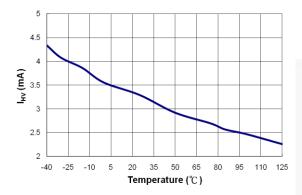


Figure 11. I<sub>HV</sub> vs. Temperature

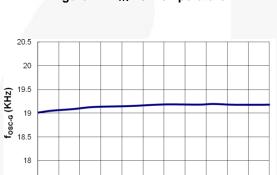


Figure 13. f<sub>OSC-G</sub> vs. Temperature

35 50 65 80

Temperature (°€)

95 110 125

-25 -10

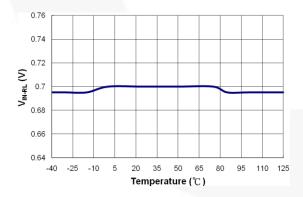


Figure 15. V<sub>IN-RL</sub> vs. Temperature

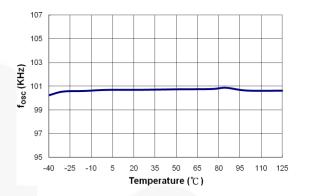


Figure 12. fosc vs. Temperature

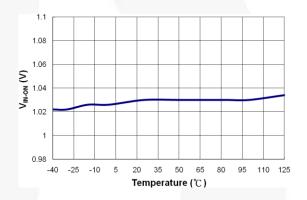


Figure 14. V<sub>IN-ON</sub> vs. Temperature

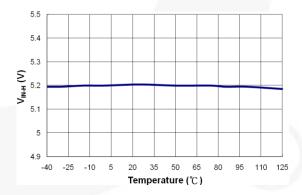
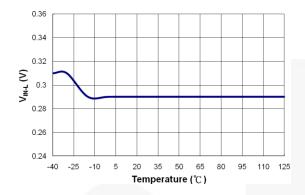


Figure 16. V<sub>IN-H</sub> vs. Temperature

### **Typical Characteristics**



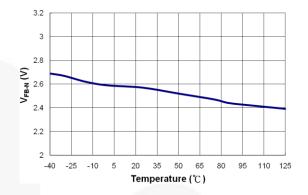
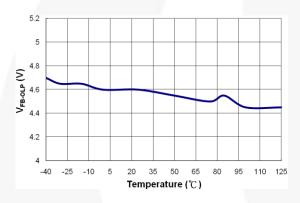


Figure 17. V<sub>IN-L</sub> vs. Temperature





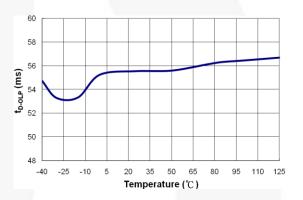
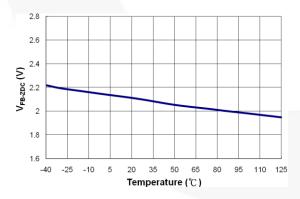


Figure 19. V<sub>FB-OLP</sub> vs. Temperature

Figure 20. t<sub>D-OLP</sub> vs. Temperature



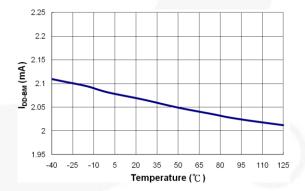


Figure 21. V<sub>FB-ZDC</sub> vs. Temperature

Figure 22. I<sub>DD-BM</sub> vs. Temperature

### **Functional Description**

### **Startup Operation**

For startup, the HV pin is connected to the line input or bulk capacitor through the external resistor,  $R_{\text{HV}}$ , as shown in Figure 23. Typical startup current drawn from the HV pin is 3.5mA and it charges the  $V_{\text{DD}}$  capacitor through the resistor  $R_{\text{HV}}$ . The startup current turns off when the  $V_{\text{DD}}$  capacitor voltage reaches  $V_{\text{DD-ON}}$ . The  $V_{\text{DD}}$  capacitor maintains  $V_{\text{DD}}$  until the auxiliary winding of the transformer provides the operating current.

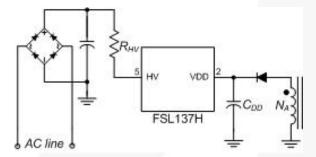


Figure 23. Startup Circuit

### **Slope Compensation**

FSL137H is designed for flyback power converters. The peak-current-mode control is used to optimize system performance. Slope compensation is added to stabilize the current loop. FSL137H inserts a synchronized, positively sloped ramp at each switching cycle.

#### Soft-Start

The FSL137H has internal soft-start circuit that slowly increases the SenseFET current after startup. The typical soft-start time is 5ms during which the  $V_{\text{Limit}}$  level is increased in six steps to smoothly establish the required output voltage, as shown in Figure 24. It also helps to prevent transformer saturation and reduce the stress on the secondary diode during startup.

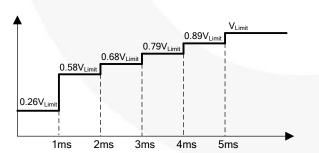


Figure 24. Soft-Start Function

#### **Green-Mode Operation**

The FSL137H uses feedback voltage (VFB) as an indicator of the output load and modulates the PWM frequency, as shown in Figure 25, such that the switching frequency decreases as load decreases. In heavy load conditions, the switching frequency is 100kHz. Once V<sub>FB</sub> decreases below V<sub>FB-N</sub> (2.5V), the PWM frequency starts to linearly decrease from 100kHz to 18kHz to reduce the switching losses. As V<sub>FB</sub> decreases below V<sub>FB-G</sub> (2.4V), the switching frequency is fixed at 18kHz and FSL137H enters into "deep" green mode to reduce the standby power consumption. As V<sub>FB</sub> decreases below V<sub>FB-ZDC</sub> (2.1V), FSL137H enters into burst-mode operation. When V<sub>FB</sub> drops below V<sub>FB</sub>. ZDC, FSL137H stops switching and the output voltage starts to drop, which causes the feedback voltage to rise. Once V<sub>FB</sub> rises above V<sub>FB-7DC</sub>, switching resumes. Burst mode alternately enables and disables switching, thereby reducing switching loss to improve power saving, as shown in Figure 26.

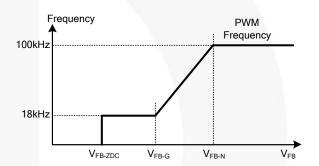


Figure 25. PWM Frequency

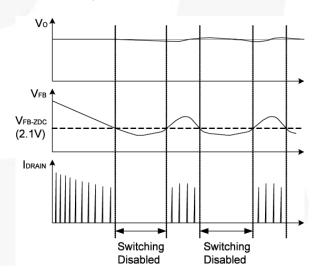


Figure 26. Burst Mode Operation

#### **Constant Power Control**

To limit the output power of the converter constantly, high/low line compensation is included. Sensing the converter input voltage through the VIN pin, the high/low line compensation function generates a relative peak-current-limit threshold voltage for constant power control, as shown in Figure 27.

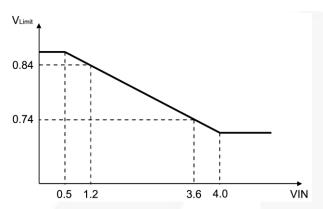


Figure 27. Constant Power Control

#### **Protections**

The FSL137H provides full protection functions to prevent the power supply and the load from being damaged. The protection features include:

#### **Latch / Auto Recovery Function**

The FSL137H provides additional protections by the VIN pin, such as pull-HIGH latch and pull-LOW auto recovery that depend on the application. As shown in Figure 28, when  $V_{\rm IN}$  is higher than 5.2V, FSL137H is latched until the  $V_{\rm DD}$  is discharged. FSL137H is in auto recovery when  $V_{\rm IN}$  is lower than 0.3V.

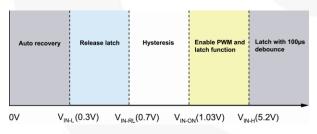
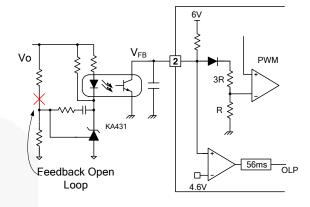


Figure 28. VIN Pin Function

#### Open-Loop / Overload Protection (OLP)

When the upper branch of the voltage divider for the shunt regulator (KA431 shown) is broken, as shown in Figure 29, or over current or output short occurs. There is no current flowing through the opto-coupler transistor, which pulls up the feedback voltage to 6V. When the feedback voltage is above 4.6V for longer than 56ms, OLP is triggered. This protection is also triggered when the SMPS output drops below the nominal value longer than 56ms due to the overload condition.



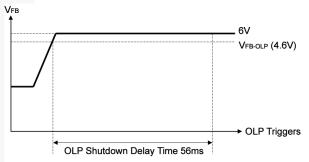


Figure 29. OLP Operation

#### **V<sub>DD</sub> Over-Voltage Protection (OVP)**

 $V_{DD}$  over-voltage protection prevents IC damage caused by over voltage on the  $V_{DD}$  pin. The OVP is triggered when  $V_{DD}$  reaches 28V. It has a debounce time (typically 130 $\mu$ s) to prevent false trigger by switching noise.

#### **Over-Temperature Protection (OTP)**

The SenseFET and the control IC are integrated, making it easier to detect the temperature of the SenseFET. When the temperature exceeds approximately 142°C, thermal shutdown is activated.

### **Typical Application Circuit**

Application	Fairchild Devices	Input Voltage Range	Output
Adapter	FSL137H	90~264Vac	12V/1A (12W)

#### **Features**

- High efficiency (>77.76% at full load) meeting Energy Star V2.0 regulation with enough margin
- Standby power <100mW at no-load condition</p>
- Provides full protection functions:

OVP	ОТР	OLP	VIN-H	VIN-L	
Latch	Latch	Auto restart	Latch	Auto restart	

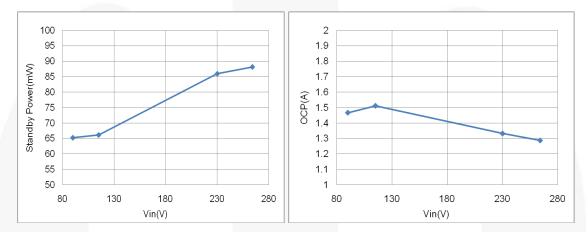


Figure 30. Measured Standby Power and OCP

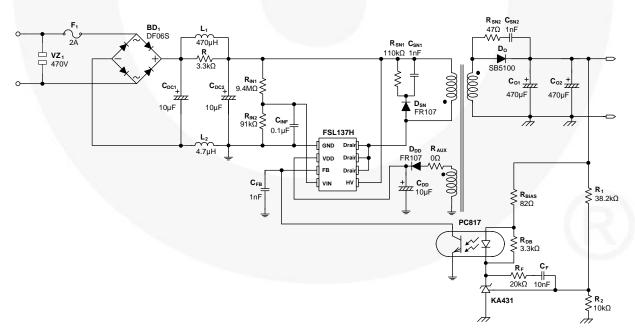


Figure 31. Schematic of Typical Application Circuit

### **Typical Application Circuit** (Continued)

### **Transformer Specification**

Core: EE16 Bobbin: EE16

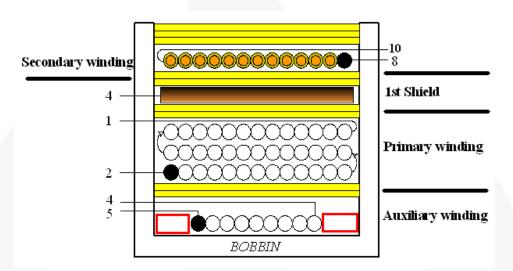
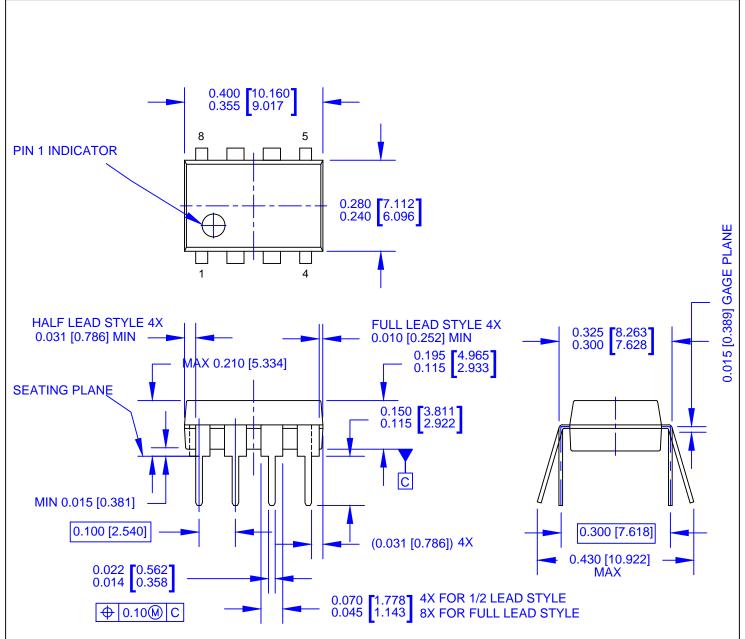


Figure 32. Transformer Diagram

NO	TERMINAL		WIRE	To				
NO.	S	F	WIKE	Ts				
W1	5	4	2UEW 0.3*1	13				
W2	2	1	2UEW 0.26*1	75				
W3	4	-	COPPER SHIELD	1.2				
W4	8	10	TEX-E 0.35*1	13				
			CORE ROUNDING TAPE	3				
Primary-Side Inductance=600µH ± 5%								
Primary-Side Effe	Primary-Side Effective Leakage<20uH + 5%							



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