# FSBB15CH60F Smart Power Module

# FAIRCHILD

SEMICONDUCTOR®

# FSBB15CH60F Smart Power Module

# Features

- UL Certified No.E209204 (SPM27-CA package)
- Very low thermal resistance due to using DBC
- 600V-15A 3-phase IGBT inverter bridge including control ICs for gate driving and protection
- Divided negative dc-link terminals for inverter current sensing applications
- Single-grounded power supply due to built-in HVIC
- Isolation rating of 2500Vrms/min.

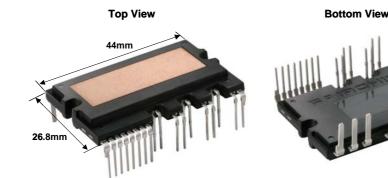
# Applications

- AC 100V ~ 253V three-phase inverter drive for small power ac motor drives
- Home appliances applications like air conditioner and washing machine.

# April 28, 2006 **SPM** ™

# **General Description**

It is an advanced smart power module (SPM<sup>TM</sup>) that Fairchild has newly developed and designed to provide very compact and high performance ac motor drives mainly targeting lowpower inverter-driven application like air conditioner and washing machine. It combines optimized circuit protection and drive matched to low-loss IGBTs. System reliability is further enhanced by the integrated under-voltage lock-out and shortcircuit protection. The high speed built-in HVIC provides optocoupler-less single-supply IGBT gate driving capability that further reduce the overall size of the inverter system design. Each phase current of inverter can be monitored separately due to the divided negative dc terminals.





# Integrated Power Functions

• 600V-15A IGBT inverter for three-phase DC/AC power conversion (Please refer to Figure 3)

# Integrated Drive, Protection and System Control Functions

- For inverter high-side IGBTs: Gate drive circuit, High voltage isolated high-speed level shifting Control circuit under-voltage (UV) protection Note) Available bootstrap circuit example is given in Figures 10 and 11.
- For inverter low-side IGBTs: Gate drive circuit, Short circuit protection (SC) Control supply circuit under-voltage (UV) protection
- Fault signaling: Corresponding to a UV fault (Low-side supply)
- Input interface: 3.3/5V CMOS/LSTTL compatible, Schmitt trigger input

# **Pin Configuration**

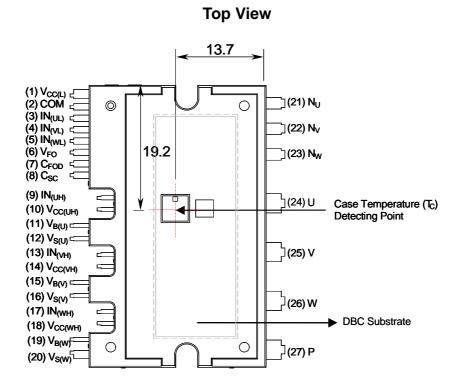
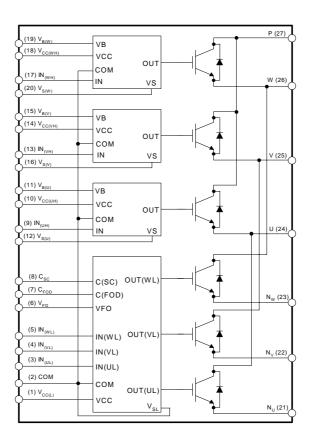


Figure 2.

Pin Number	Pin Name	Pin Description		
1	V <sub>CC(L)</sub>	Low-side Common Bias Voltage for IC and IGBTs Driving		
2	COM	Common Supply Ground		
3	IN <sub>(UL)</sub>	Signal Input for Low-side U Phase		
4	IN <sub>(VL)</sub>	Signal Input for Low-side V Phase		
5	IN <sub>(WL)</sub>	Signal Input for Low-side W Phase		
6	V <sub>FO</sub>	Fault Output		
7	C <sub>FOD</sub>	Capacitor for Fault Output Duration Time Selection		
8	C <sub>SC</sub>	Capacitor (Low-pass Filter) for Short-Current Detection Input		
9	IN <sub>(UH)</sub>	Signal Input for High-side U Phase		
10	V <sub>CC(UH)</sub>	High-side Bias Voltage for U Phase IC		
11	V <sub>B(U)</sub>	High-side Bias Voltage for U Phase IGBT Driving		
12	V <sub>S(U)</sub>	High-side Bias Voltage Ground for U Phase IGBT Driving		
13	IN <sub>(VH)</sub>	Signal Input for High-side V Phase		
14	V <sub>CC(VH)</sub>	High-side Bias Voltage for V Phase IC		
15	V <sub>B(V)</sub>	High-side Bias Voltage for V Phase IGBT Driving		
16	V <sub>S(V)</sub>	High-side Bias Voltage Ground for V Phase IGBT Driving		
17	IN <sub>(WH)</sub>	Signal Input for High-side W Phase		
18	V <sub>CC(WH)</sub>	High-side Bias Voltage for W Phase IC		
19	V <sub>B(W)</sub>	High-side Bias Voltage for W Phase IGBT Driving		
20	V <sub>S(W)</sub>	High-side Bias Voltage Ground for W Phase IGBT Driving		
21	NU	Negative DC-Link Input for U Phase		
22	N <sub>V</sub>	Negative DC-Link Input for V Phase		
23	N <sub>W</sub>	Negative DC-Link Input for W Phase		
24	U	Output for U Phase		
25	V	Output for V Phase		
26	W	Output for W Phase		
27	Р	Positive DC–Link Input		

# Internal Equivalent Circuit and Input/Output Pins



### Note:

1. Inverter low-side is composed of three IGBTs, freewheeling diodes for each IGBT and one control IC. It has gate drive and protection functions.

2. Inverter power side is composed of four inverter dc-link input terminals and three inverter output terminals.

3. Inverter high-side is composed of three IGBTs, freewheeling diodes and three drive ICs for each IGBT.



# Absolute Maximum Ratings (T<sub>J</sub> = 25°C, Unless Otherwise Specified)

### **Inverter Part**

Symbol	Parameter	Conditions	Rating	Units
V <sub>PN</sub>	Supply Voltage	Applied between P- N <sub>U</sub> , N <sub>V</sub> , N <sub>W</sub>	450	V
V <sub>PN(Surge)</sub>	Supply Voltage (Surge)	Applied between P- $N_U$ , $N_V$ , $N_W$	500	V
V <sub>CES</sub>	Collector-emitter Voltage		600	V
± I <sub>C</sub>	Each IGBT Collector Current	$T_{\rm C} = 25^{\circ}{\rm C}$	15	А
± I <sub>CP</sub>	Each IGBT Collector Current (Peak)	$T_{C} = 25^{\circ}C$ , Under 1ms Pulse Width	30	А
P <sub>C</sub>	Collector Dissipation	T <sub>C</sub> = 25°C per One Chip	50	W
ТJ	Operating Junction Temperature	(Note 1)	-20 ~ 125	°C

Note:

1. The maximum junction temperature rating of the power chips integrated within the SPM is 150 °C(@T<sub>C</sub> ≤ 100°C). However, to insure safe operation of the SPM, the average junction temperature should be limited to T<sub>J(ave)</sub> ≤ 125°C (@T<sub>C</sub> ≤ 100°C)

# **Control Part**

Symbol	Parameter Conditions		Rating	Units
V <sub>CC</sub>	Control Supply Voltage	Applied between V <sub>CC(UH)</sub> , V <sub>CC(VH)</sub> , V <sub>CC(WH)</sub> , V <sub>CC(L)</sub> - COM	20	V
V <sub>BS</sub>	High-side Control Bias Voltage	Applied between $V_{B(U)}$ - $V_{S(U)},\ V_{B(V)}$ - $V_{S(V)},\ V_{B(W)}$ - $V_{S(W)}$	20	V
V <sub>IN</sub>	Input Signal Voltage	Applied between $\rm IN_{(UH)},~\rm IN_{(VH)},~\rm IN_{(WH)},~\rm IN_{(UL)},~\rm IN_{(VL)},~\rm IN_{(VL)},~\rm IN_{(WL)}$	-0.3~17	V
V <sub>FO</sub>	Fault Output Supply Voltage	Applied between V <sub>FO</sub> - COM	-0.3~V <sub>CC</sub> +0.3	V
I <sub>FO</sub>	Fault Output Current	Sink Current at V <sub>FO</sub> Pin	5	mA
V <sub>SC</sub>	Current Sensing Input Voltage	Applied between C <sub>SC</sub> - COM	-0.3~V <sub>CC</sub> +0.3	V

# **Total System**

Symbol	Parameter	Conditions	Rating	Units
V <sub>PN(PROT)</sub>	Self Protection Supply Voltage Limit (Short Circuit Protection Capability)	$V_{CC} = V_{BS} = 13.5 \sim 16.5 V$ T <sub>J</sub> = 125°C, Non-repetitive, less than 2µs	400	V
T <sub>C</sub>	Module Case Operation Temperature	-20°C $\leq$ T_J $\leq$ 125°C, See Figure 2	-20 ~ 100	°C
T <sub>STG</sub>	Storage Temperature		-40 ~ 125	°C
V <sub>ISO</sub>	Isolation Voltage	60Hz, Sinusoidal, AC 1 minute, Connection Pins to DBC substrate	2500	V <sub>rms</sub>

# **Thermal Resistance**

Symbol	Parameter	Conditions		Тур.	Max.	Units
R <sub>th(j-c)Q</sub>	Junction to Case Thermal	Inverter IGBT part (per 1/6 module)	-	-	2.02	°C/W
R <sub>th(j-c)F</sub>	Resistance	Inverter FWD part (per 1/6 module)	-	-	3.15	°C/W

Note:

2. For the measurement point of case temperature(T\_C), please refer to Figure 2.

# Package Marking and Ordering Information

Device Marking	Device	Package	Reel Size	Tape Width	Quantity
FSBB15CH60F	FSBB15CH60F	SPM27CA	-	-	10

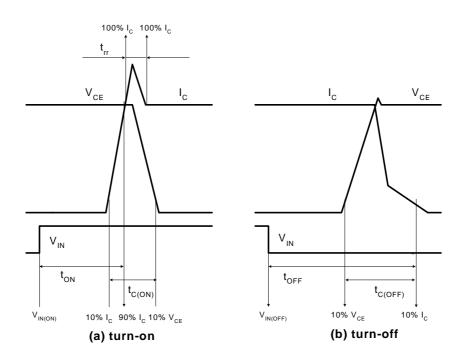
# Electrical Characteristics (T<sub>J</sub> = 25°C, Unless Otherwise Specified)

### **Inverter Part**

Sy	ymbol	Parameter	Condi	tions	Min.	Тур.	Max.	Units
Vo	CE(SAT)	Collector-Emitter Saturation Voltage	$V_{CC} = V_{BS} = 15V$ $V_{IN} = 5V$	I <sub>C</sub> =15A, T <sub>J</sub> = 25°C	-	-	2.3	V
	V <sub>F</sub>	FWD Forward Voltage	$V_{IN} = 0V$	I <sub>C</sub> = 15A, T <sub>J</sub> = 25°C	-	-	2.1	V
HS	t <sub>ON</sub>	Switching Times	$V_{PN} = 300V, V_{CC} = V_{BS} = 15V$		-	0.4	-	μS
	t <sub>C(ON)</sub>		$I_{C} = 15A$ $V_{IN} = 0V \leftrightarrow 5V$ , Inducti	veload	-	0.28	-	μS
	t <sub>OFF</sub>		(Note 3) $(\text{Note 3})$		-	0.67	-	μS
	t <sub>C(OFF)</sub>				-	0.35	-	μS
	t <sub>rr</sub>				-	0.10	-	μS
LS	t <sub>ON</sub>		$V_{PN} = 300V, V_{CC} = V_{BS}$	<sub>S</sub> = 15V	-	0.55	-	μS
	t <sub>C(ON)</sub>		$I_{C} = 15A$ $V_{IN} = 0V \leftrightarrow 5V$ , Inducti	veload	-	0.24	-	μS
	t <sub>OFF</sub>		(Note 3) $(\text{Note 3})$		-	0.73	-	μS
	t <sub>C(OFF)</sub>				-	0.34	-	μS
	t <sub>rr</sub>	1			-	0.10	-	μS
	I <sub>CES</sub>	Collector-Emitter Leakage Current	V <sub>CE</sub> = V <sub>CES</sub>		-	-	250	μΑ

Note:

3. t<sub>ON</sub> and t<sub>OFF</sub> include the propagation delay time of the internal drive IC. t<sub>C(ON)</sub> and t<sub>C(OFF)</sub> are the switching time of IGBT itself under the given gate driving condition internally. For the detailed information, please see Figure 4.



# Figure 4. Switching Time Definition

# **Electrical Characteristics** ( $T_J = 25^{\circ}C$ , Unless Otherwise Specified)

# **Control Part**

Symbol	Parameter	Со	nditions	Min.	Тур.	Max.	Units
I <sub>QCCL</sub>	Quiescent V <sub>CC</sub> Supply Current	V <sub>CC</sub> = 15V IN <sub>(UL, VL, WL)</sub> = 0V	V <sub>CC(L)</sub> - COM	-	-	23	mA
I <sub>QCCH</sub>		V <sub>CC</sub> = 15V IN <sub>(UH, VH, WH)</sub> = 0V	$\begin{array}{l} V_{CC(UH)},  V_{CC(VH)}, \\ V_{CC(WH)} \ \text{-} \ COM \end{array}$	-	-	100	μΑ
I <sub>QBS</sub>	Quiescent V <sub>BS</sub> Supply Current	V <sub>BS</sub> = 15V IN <sub>(UH, VH, WH)</sub> = 0V	$ \begin{array}{l} V_{B(U)} \mbox{-} V_{S(U)}, V_{B(V)} \mbox{-} V_{S(V)}, \\ V_{B(W)} \mbox{-} V_{S(W)} \end{array} $	-	-	500	μΑ
V <sub>FOH</sub>	Fault Output Voltage	$V_{SC} = 0V, V_{FO}$ Circuit: 4.7k $\Omega$ to 5V Pull-up		4.5	-	-	V
V <sub>FOL</sub>		V <sub>SC</sub> = 1V, V <sub>FO</sub> Circu	$V_{SC}$ = 1V, $V_{FO}$ Circuit: 4.7k $\Omega$ to 5V Pull-up		-	0.8	V
V <sub>SC(ref)</sub>	Short Circuit Trip Level	V <sub>CC</sub> = 15V (Note 4)		0.45	0.5	0.55	V
TSD	Over-temperature protec- tion	Temperature at LVIC		-	145	-	°C
∆TSD	Over-temperature protec- tion hysterisis	Temperature at LVIC		-	18	-	°C
UV <sub>CCD</sub>	Supply Circuit Under-	Detection Level		10.7	11.9	13.0	V
UV <sub>CCR</sub>	Voltage Protection	Reset Level		11.2	12.4	13.2	V
UV <sub>BSD</sub>		Detection Level		10.1	11.3	12.5	V
UV <sub>BSR</sub>		Reset Level		10.5	11.7	12.9	V
t <sub>FOD</sub>	Fault-out Pulse Width	C <sub>FOD</sub> = 33nF (Note 5)		1.0	1.8	-	ms
V <sub>IN(ON)</sub>	ON Threshold Voltage	Applied between IN <sub>(UH)</sub> , IN <sub>(VH)</sub> , IN <sub>(WH)</sub> , IN <sub>(UL)</sub> ,		3.0	-	-	V
V <sub>IN(OFF)</sub>	OFF Threshold Voltage	IN <sub>(VL)</sub> , IN <sub>(WL)</sub> - COM		-	-	0.8	V

Note:

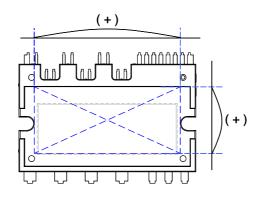
4. Short-circuit current protection is functioning only at the low-sides.

5. The fault-out pulse width  $t_{FOD}$  depends on the capacitance value of  $C_{FOD}$  according to the following approximate equation :  $C_{FOD} = 18.3 \times 10^{-6} \times t_{FOD}[F]$ 

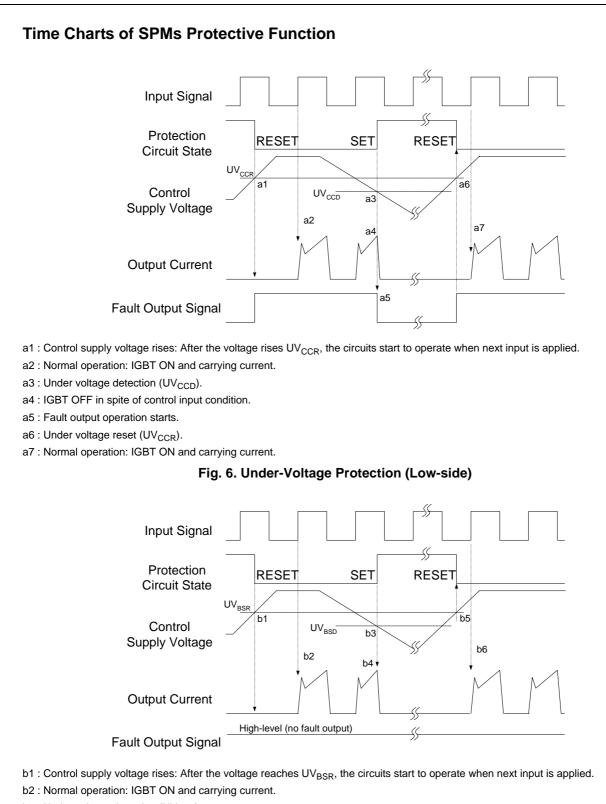
# **Recommended Operating Conditions**

Symbol	Parameter	Conditions	Value			Units
Symbol Parameter		Conditions	Min.	Тур.	Max.	Units
V <sub>PN</sub>	Supply Voltage	Applied between P - N <sub>U</sub> , N <sub>V</sub> , N <sub>W</sub>	-	300	400	V
V <sub>CC</sub>	Control Supply Voltage	Applied between V <sub>CC(UH)</sub> , V <sub>CC(VH)</sub> , V <sub>CC(WH)</sub> , V <sub>CC(WH)</sub> , V <sub>CC(L)</sub> - COM	13.5	15	16.5	V
V <sub>BS</sub>	High-side Bias Voltage	Applied between V <sub>B(U)</sub> - V <sub>S(U)</sub> , V <sub>B(V)</sub> - V <sub>S(V)</sub> , V <sub>B(W)</sub> - V <sub>S(W)</sub>		15	18.5	V
DV <sub>CC</sub> /Dt, DV <sub>BS</sub> /Dt	Control supply variation		-1	-	1	V/µs
t <sub>dead</sub>	Blanking Time for Preventing Arm-short	nking Time for Preventing For Each Input Signal		-	-	μS
f <sub>PWM</sub>	PWM Input Signal $-20^{\circ}C \le T_C \le 100^{\circ}C, -20^{\circ}C \le T_J \le 125^{\circ}C$		-	-	20	kHz
V <sub>SEN</sub>	Voltage for Current Sensing	Applied between N <sub>U</sub> , N <sub>V</sub> , N <sub>W</sub> - COM (Including surge voltage)	-4		4	V

Doromotor		Condition			Limits			
Parameter		onation	Min.	Тур.	Max.	Units		
Mounting Torque	Mounting Screw: - M3	Recommended 0.62N•m	0.51	0.62	0.72	N∙m		
Device Flatness		Note Figure 5	0	-	+120	μm		
Weight			-	15.00	-	g		

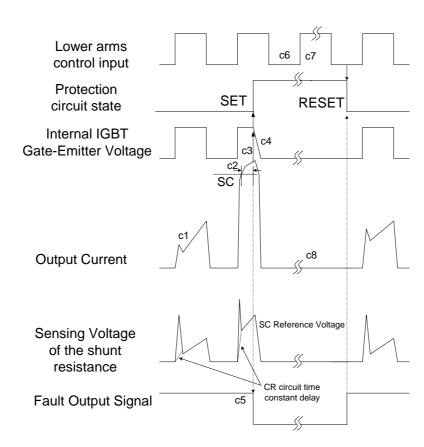






- b3 : Under voltage detection (UV<sub>BSD</sub>).
- b4 : IGBT OFF in spite of control input condition, but there is no fault output signal.
- b5 : Under voltage reset (UV<sub>BSR</sub>)
- b6 : Normal operation: IGBT ON and carrying current

# Fig. 7. Under-Voltage Protection (High-side)



(with the external shunt resistance and CR connection)

c1 : Normal operation: IGBT ON and carrying current.

c2 : Short circuit current detection (SC trigger).

c3 : Hard IGBT gate interrupt.

c4 : IGBT turns OFF.

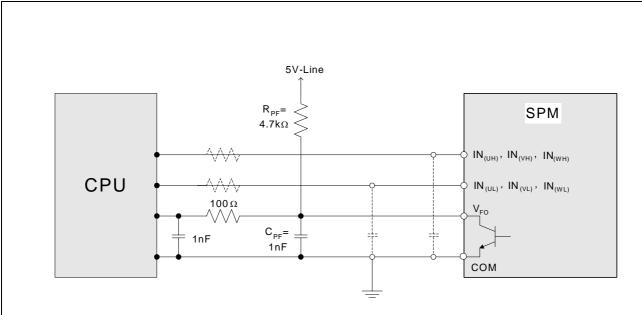
c5 : Fault output timer operation starts: The pulse width of the fault output signal is set by the external capacitor  $C_{FO}$ .

c6 : Input "L" : IGBT OFF state.

c7 : Input "H": IGBT ON state, but during the active period of fault output the IGBT doesn't turn ON.

c8 : IGBT OFF state

# Fig. 8. Short-Circuit Current Protection (Low-side Operation only)

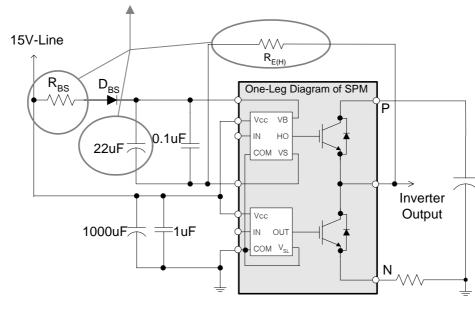


### Note:

1. RC coupling at each input (parts shown dotted) might change depending on the PWM control scheme used in the application and the wiring impedance of the application's printed circuit board. The SPM input signal section integrates 3.3kΩ(typ.) pull-down resistor. Therefore, when using an external filtering resistor, please pay attention to the signal voltage drop at input terminal.

2. The logic input is compatible with standard CMOS or LSTTL outputs.

### Fig. 9. Recommended CPU I/O Interface Circuit



# These Values depend on PWM Control Algorithm

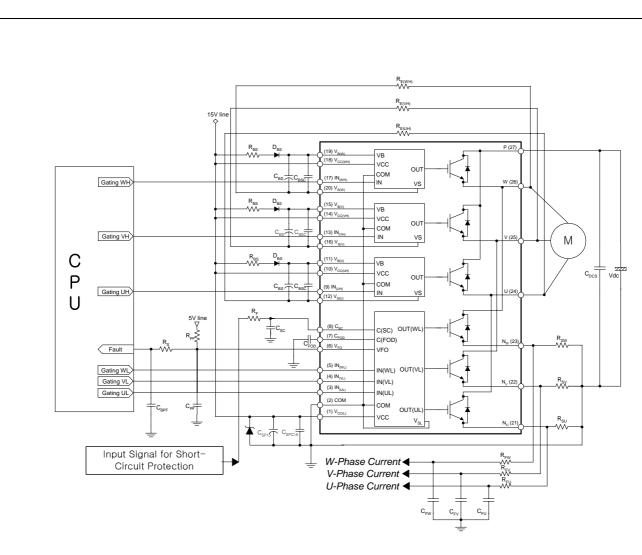
### Note:

1. It would be recommended that the bootstrap diode,  $\mathsf{D}_{\mathsf{BS}}$ , has soft and fast recovery characteristics.

2. The bootstrap resistor (R<sub>BS</sub>) should be 3 times greater than R<sub>E(H)</sub>. The recommended value of R<sub>E(H)</sub> is 5.6Ω, but it can be increased up to 20Ω (maximum) for a slower dv/dt of high-side.

3. The ceramic capacitor placed between  $V_{CC}$ -COM should be over 1µF and mounted as close to the pins of the SPM as possible.

# Fig. 10. Recommended Bootstrap Operation Circuit and Parameters

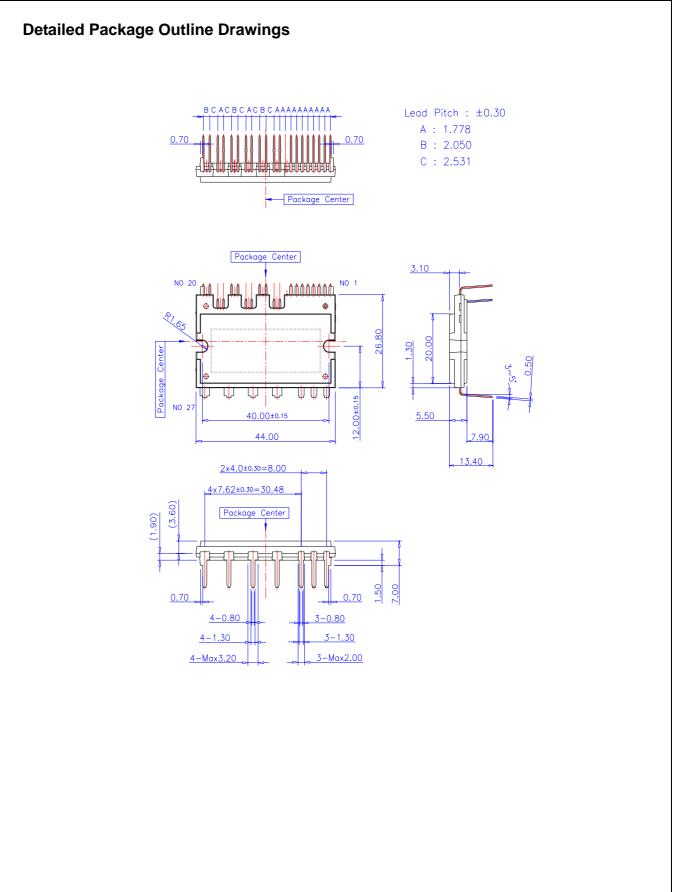


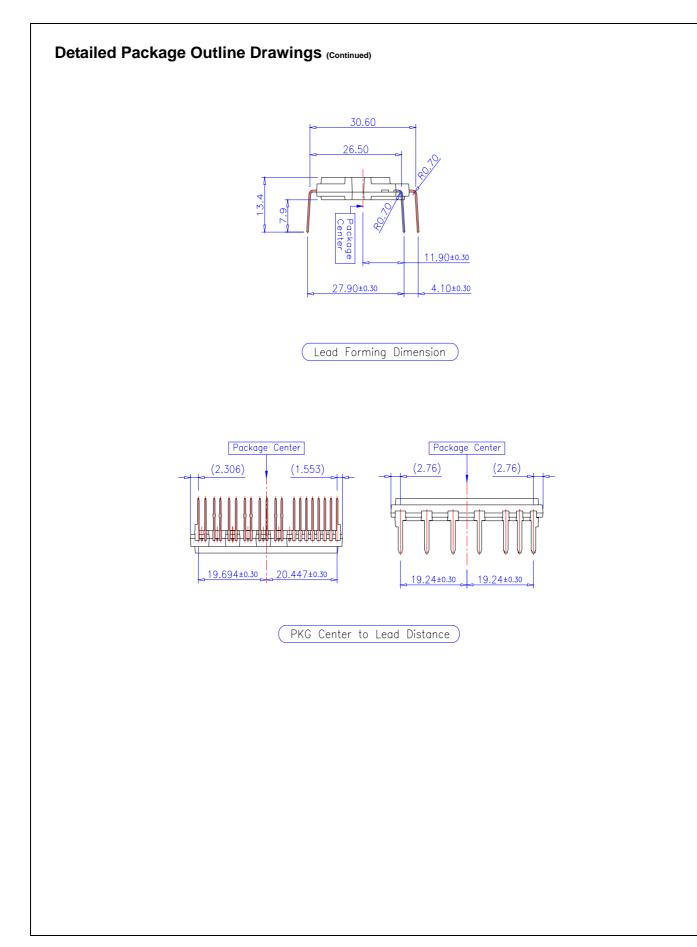
### Note:

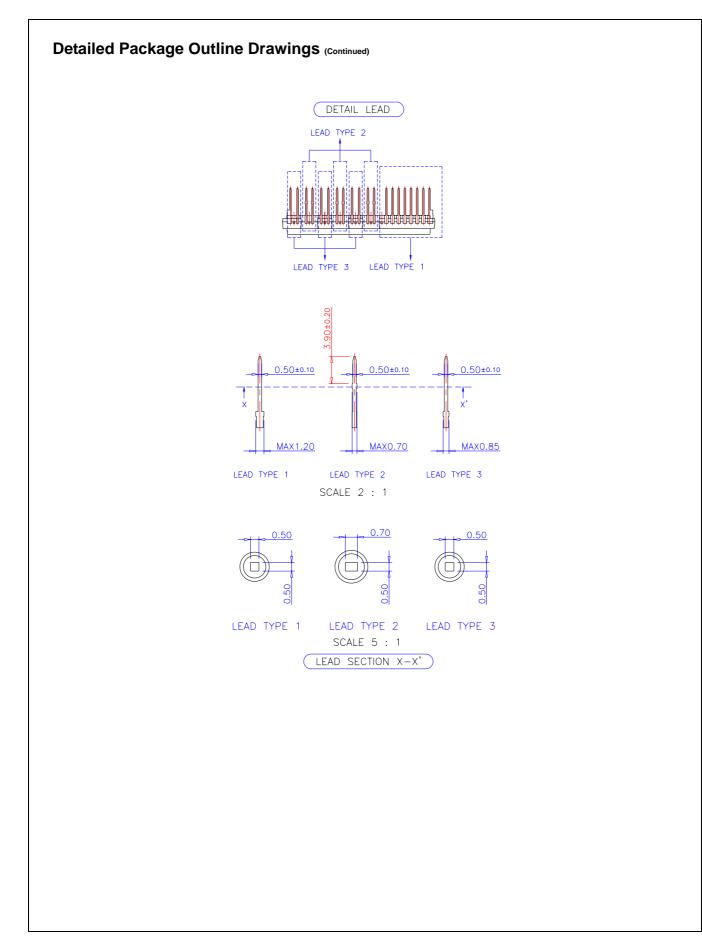
- 1. To avoid malfunction, the wiring of each input should be as short as possible. (less than 2-3cm)
- 2. By virtue of integrating an application specific type HVIC inside the SPM, direct coupling to CPU terminals without any opto-coupler or transformer isolation is possible.
- 3.  $V_{FO}$  output is open collector type. This signal line should be pulled up to the positive side of the 5V power supply with approximately 4.7k $\Omega$  resistance. Please refer to Figure. 9. 4.  $C_{SP15}$  of around 7 times larger than bootstrap capacitor  $C_{BS}$  is recommended.
- 5. V<sub>FO</sub> output pulse width should be determined by connecting an external capacitor(C<sub>FOD</sub>) between C<sub>FOD</sub>(pin7) and COM(pin2). (Example : if C<sub>FOD</sub> = 33 nF, then t<sub>FO</sub> = 1.8ms (typ.)) Please refer to the note 5 for calculation method.
- 6. Input signal is High-Active type. There is a 3.3kΩ resistor inside the IC to pull down each input signal line to GND. When employing RC coupling circuits, set up such RC couple that input signal agree with turm-off/turm-on threshold voltage.
- 7. To prevent errors of the protection function, the wiring around  ${\sf R}_{\sf F}$  and  ${\sf C}_{\sf SC}$  should be as short as possible.
- 8. In the short-circuit protection circuit, please select the  $R_FC_{SC}$  time constant in the range 1.5~2  $\mu s.$
- 9. Each capacitor should be mounted as close to the pins of the SPM as possible.
- 10. To prevent surge destruction, the wiring between the smoothing capacitor and the P&GND pins should be as short as possible. The use of a high frequency non-inductive capacitor of around 0.1~0.22 μF between the P&GND pins is recommended.
- 11. Relays are used at almost every systems of electrical equipments of home appliances. In these cases, there should be sufficient distance between the CPU and the relays.

12.  $C_{\mbox{SPC15}}$  should be over 1uF and mounted as close to the pins of the SPM as possible.

# Fig. 11. Typical Application Circuit









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Preliminary	First Production	This datasheet contains preliminary data; supplementary data will be published at a later date. Fairchild Semiconductor reserves the right to make changes at any time without notice to improve design.
No Identification Needed	Full Production	This datasheet contains final specifications. Fairchild Semiconductor reserves the right to make changes at any time without notice to improve design.
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