FPD33684A/ FPD33684B Low Power, Low EMI, TFT-LCD Column Driver with RSDS™ Inputs, 64 Grayshades, and 384 Outputs for XGA/SXGA Applications



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FPD33684A/ FPD33684B Low Power, Low EMI, TFT-LCD Column Driver with **RSDS™** Inputs, 64 Grayshades, and 384 Outputs for XGA/SXGA Applications

General Description

The FPD33684 Column Driver is a direct drive, 64 gray level, 384 output, TFT-LCD column driver with an RSDS[™] data interface. It provides the capability to display 262.144 colors (18-bit color) with a large dynamic output range for twisted nematic applications. When used in a bank with other FPD33684 column drivers, the FPD33684 can support both XGA (8 drivers) or SXGA (10 drivers) applications. Output voltages are programmably gamma corrected to provide a direct mapping between digital video and LCD panel brightness.

An RSDS[™] (Reduced Swing Differential Signaling) interface is used between the timing controller and the column driver to minimize EMI and reduce power.

The FPD33684 offers a low power, low EMI column driver solution with direct-drive dynamic range and dot-inversion addressing.

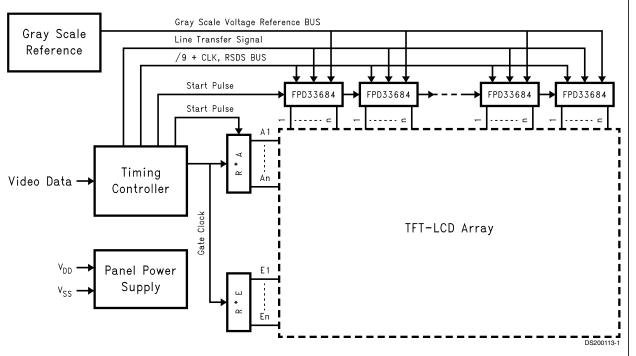
Features

- RSDS[™] (Reduced Swing Differential Signaling) data bus for low power, reduced EMI and small PCB foot print
- Up to 85MHz clock
- Supports both XGA and SXGA timing
- Supports notebook and monitor applications
- Smart Charge Conservation for low power consumption
- 64 Gray levels per color (18-bit color)
- Supports both Dot and N-Line inversion
- Externally programmable gamma characteristic
- Very low offsets for artifact-free images
- High voltage outputs for high contrast in a large range of display panel applications
- Optional, high current, repair line buffers
- Available in 2 common gamma reference curves

Ordering Information			
Part Number	Gamma Curve	Custom TCP #	Package Suffix
FPD33684	A or B	XX(Note 1)	СТ

Note 1: Custom TCP # is assigned by National Semiconductor for each custom TCP design

System Diagram



Absolute Maximum Ratings (Note 2)

Analog Supply, (V _{DD2}) (Note 3)	-0.3V to +11.5V
Logic Supply, (V _{DD1}) (Note 3)	-0.3V to +5.0V
High Bias Supply, (V _{HBIAS}) (Note 3)	-0.3V to +13.0V
Low-Polarity RDAC Reference	
Voltages, (V _{GMA6} to V _{GMA10}) (Note 3)	$-0.3V$ to $0.5V_{\text{DD2}}$
High-Polarity RDAC Reference Voltages, (V _{GMA1} to V _{GMA5}) (Note 3)	$0.5V_{DD2} - 1.0V$ to $V_{DD2} + 0.3V$
RDAC Current (All Gamma Voltage Taps), (I_{GMA} to I_{GMA10})	-2.5mA to 2.5mA
Input Voltage (Digital Logic), (V _{IN}) (Note 3)	–0.3V to V _{DD1} + 0.3V
Output Voltage, (V _{OUT}) (Note 3)	–0.3V to V_{DD2} + 0.3V
Output Current (Analog), (I _{OUT})	–7mA to +7mA
Storage Temperature Range, (T_S)	–55°C to +125°C

Note 2: "Absolute Maximum Ratings" are those values beyond which the safety of the device cannot be guaranteed. They are not meant to imply that the devices should be operated at these limits. The table of "Electrical Characteristics" specifies conditions of device operation.

Note 3: Absolute voltages referenced to $V_{SS1} = V_{SS2} = 0.0V$.

Recommended Operating Conditions

	Min	Тур	Max	Units
Logic Supply Voltage (V _{DD1})	2.7	3.3	3.6	V
Supply Voltage (V _{DD2})	7.5		10.5	V
Supply Voltage (V _{HBIAS})	V_{DD2}	,	V _{DD2} +1.5	V
Operating Temperature				
(T _A)	-10	+25	+70	°C

DC Electrical Characteristics Digital Electrical Characteristics

Symbol Parameter Conditions Min Тур Max Units $V_{\rm IH}$ Logic Input High Voltage 0.7 V V_{DD1} VIL Logic Input Low Voltage 0.3 V V_{DD1} V_{OH} $I_{OH} = -0.5 m \overline{A}$ Logic Output High Voltage V_{DD1} – V 0.5 V_{OL} 0.5 V Logic Output Low Voltage $I_{OL} = 0.5 mA$ (Note 4) Logic Current 8.0 3.0 mA I_{DD1} Input Leakage $V_{DD1} = 3.6V, V_{IN} = 3.6V$ 1 $I_{\rm H}$ -1 μΑ $V_{DD1} = 3.6V, V_{IN} = 0V$ 1 Input Leakage -1 $|_{|L}$ μΑ CIN Input Capacitance All logic pins 2 pF

Note 4: CLK frequency = 32.5 MHz, V_{DD1} = 3.3V, V_{SS1} = V_{SS2} = 0.0V, charge share time = 1.5µs, line time = 22µs.

RSDS Characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Units
VIH _{RSDS}	RSDS™ High Input Voltage	VCM _{RSDS} = 1.2V (Note 5) see Figure 1	100	200		mV
VIL _{RSDS}	RSDS [™] Low Input Voltage	VCM _{RSDS} = 1.2V (Note 5) see Figure 1		-200	-100	mV
VCM _{RSDS}	RSDS™ Common Mode Input Voltage Range	VIH _{RSDS} = +100mV, VIL _{RSDS} = -100mV (Note 6) see <i>Figure 1</i>	V _{SS1} + 0.1		V _{DD1} – 1.3	V
IDL	RSDS [™] Input Leakage Current	DxxP, DxxN, CLKP, CLKN	-10		10	μA

Note 5: VCM_{RSDS} = (VCLKP + VCLKN)/2 or (VDxxP + VDxxN)/2.

Note 6: Positive means that DxxP (or CLKP) is higher than RSDS ground DxxN (or CLKN). Negative means that DxxP (or CLKP) is lower than RSDS ground DxxN (or CLKN).

RSDS Characteristics (Continued)



Analog Electrical Characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Units
I _{DD2}	Supply Current Consumption	(Note 7)		3.0	8.0	mA
I _{HBIAS}	Current Consumption through HBIAS pin			1.25		mA
PD	Power Dissipation	(Note 7)		45		mW
V_{GMA1}	Upper RDAC High Side Input	(Note 8)	V _{DD2} /2 + 0.2		V _{DD2} – 0.2	V
V_{GMA5}	Upper RDAC Low Side Input	(Note 8)	V _{DD2} /2 + 0.2		V _{DD2} – 0.2	V
V_{GMA6}	Lower RDAC High Side Input	(Note 8)	0.2		V _{DD2} /2 - 0.2	V
V _{GMA10}	Lower RDAC Low Side Input	(Note 8)	0.2		V _{DD2} /2 - 0.2	V
V _{CS}	Charge Share Voltage		The Greater of V _{DD1} or V _{GMA6}		V _{GMA5}	V
C _{LOAD}	Output Capacitive Load		30		150	pF
V _{OUT}	Output Voltage Range		V _{SS2} + 0.2		V _{DD2} – 0.2	V
R _{DAC}	RDAC References (V_{GMA1} to V_{GMA5} and V_{GMA6} to V_{GMA10})	each	12.0	15.0	18.0	kΩ
V _{pperr}	Output Peak to Peak Error (gray levels 0 through 58)	$V_{GMA1} = V_{DD2} - 0.2V$ $V_{GMA10} = V_{SS2} + 0.2V$		±3	±12	mV
	Output Peak to Peak Error (gray levels 59 through 63)	(Note 9)		±5	±25	mV
V _{parterr}	Output Part to Part Error	(Note 10)			±5	mV
I _{OUT RP}	Repair Buffer Output Current	(Note 11)	±2	±3		mA

Note 7: $V_{DD2} = 10.5V$, $V_{HBIAS} = 10.5V$, $V_{DD1} = 3.3V$, DCLK = 65 MHz, $R_{LOAD} = 5 \text{ k}\Omega$, $C_{LOAD} = 50 \text{ pF}$, charge share time = 1.5 μ s, all other swinging between V_{GMA1} (= 8.0V) and V_{GMA10} (= 0.5V) with a line time = 22 μ s.

Note 8: The following relationship must be maintained between the reference voltages: $V_{DD2} > V_{GMA1} > V_{GMA2} > V_{GMA3} > V_{GMA4} > V_{GMA5} > V_{GMA6} > V_{GMA7} > V_{GMA8} > V_{GMA8} > V_{GMA9} > V_{GMA1} > V_{SS2}$

Note 9: V_{pperr} is defined as the error in peak-to-peak output voltage for each gray level when the output swings from the gray level high value (VHxx) to the gray level low value (VLxx). This parameter applies to every output on the die. The typical value represents one standard deviation from ideal based on final test data. **Note 10:** $V_{parterr}$ is meant to guarantee the part to part output variation. The average of all outputs at gray level 32 is compared to a nominal gray level 32 value. **Note 11:** Current into device pins is defined as positive. Current out of device pins is defined as negative. $|V_{OUT} - V_{IN}| > 500$ mV.

AC Electrical Characteristics Digital AC Characteristics

Symbol	Parameter	Conditions	Min	Тур	Max	Units
PW _{CLK}	Clock Period		11.7			ns
PW _{CLK(L)}	Low Clock Pulse Width		5			ns
PW _{CLK(H)}	High Clock Pulse Width		5			ns
t _{setup1}	RSDS Data Setup Time		2			ns
t _{hold1}	RSDS Data Hold Time		0			ns
t _{setup2}	ENIOx Setup Time		2			ns
t _{hold2}	ENIOx Hold Time		4			ns
t _{PLH1}	Start Pulse Fall Delay	C _{LINE} = 15 pF			8	ns
PW _{DIO}	ENIOx Pulse Width		1		2	ТСЬК
PW _{CLK1}	LOAD Pulse Width		5 T _{CLK}		5µs	
t _{LDT}	Last Clock to LOAD Delay		5			T _{CLK}
t _{DENSU}	LOAD to First ENIO Setup		2			T _{CLK}
tPOL-CLK1	POL-CLK1 Time		14			ns

Analog AC Characteristics Supplies: $V_{SS1} = V_{SS2} = 0.0V$, $V_{DD1} = 3.3V$, $V_{DD2} = +9.5V$, $V_{HBIAS} = 11.0V$.

Symbol	Parameter	Conditions	Min	Тур	Max	Units
t _{settle 90%}	Output Settling Time to 90% of Final Value	Figure 2 (Note 12)			6	μs
t _{6-bit accy}	Output Settling Time to 6-bit accuracy	Figure 2 (Note 12)			10	μs
t _{RP 90%}	Repair Line Output Settling Time to 90% of Final Value	C _{LOAD} = 150 pF, (Note 12)			6	μs
t _{RP 6-bit accy}	Repair Line Output Settling Time to 6-bit accuracy	C _{LOAD} = 150 pF, (Note 12)			10	μs

Note 12: Charge Share Time = 800ns, V_{GMA1} = 10.3V, V_{GMA10} = 0.2V, V_{GMA5} = 5.45V, V_{GMA6} = 5.05V.

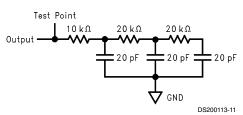
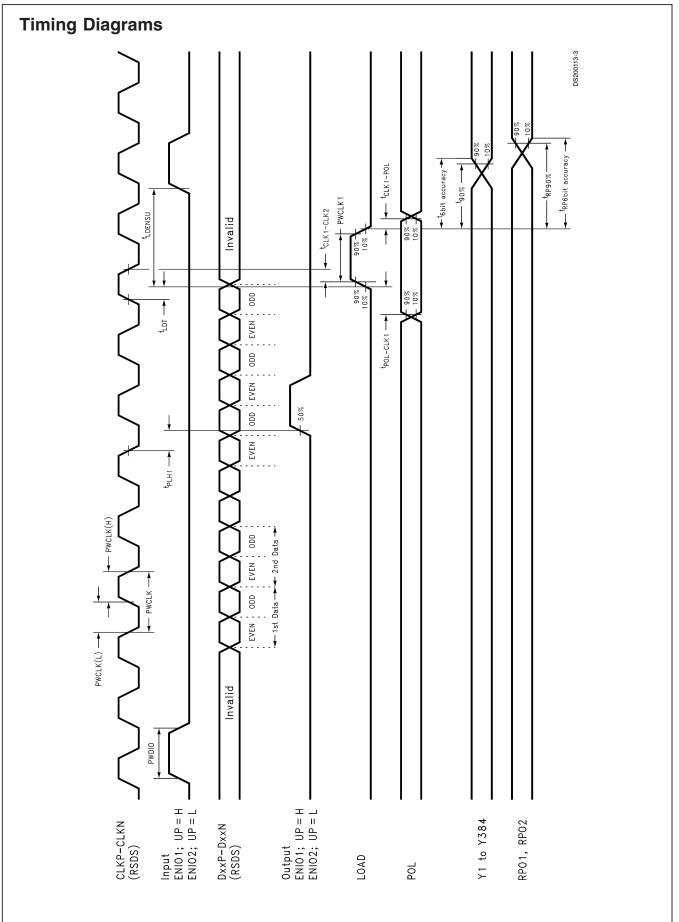
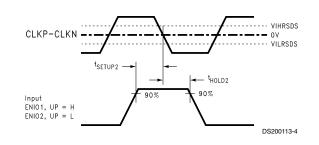


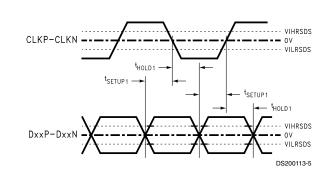
FIGURE 2. Test Circuit for Output Settling Time Measurements

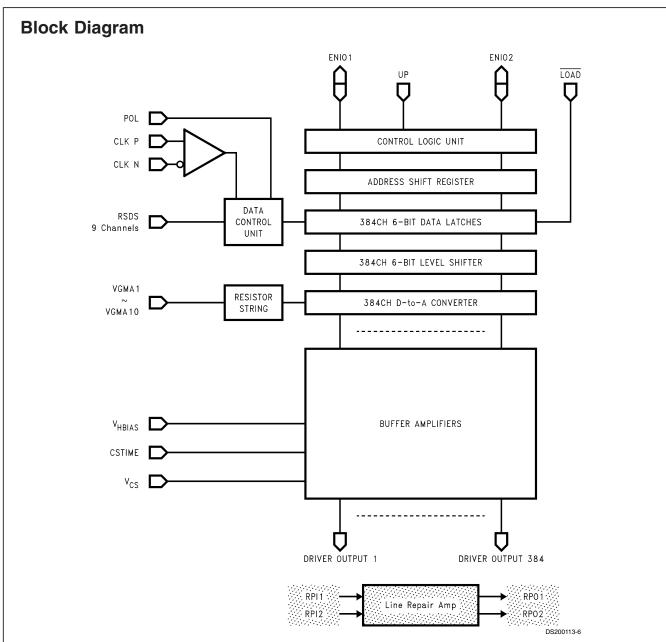


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Timing Diagrams (Continued)







Functional Description

GENERAL OVERVIEW

The FPD33684 is a low power, low EMI, 384 output column driver with 64 gray level capability (6-bit). It provides direct drive for TFT-LCD displays, eliminating the need for V_{com} modulation. Direct drive significantly reduces system power consumption and also reduces component count while providing superior image quality and cross-talk margin. The FPD33684 utilizes National's *Charge Conservation Technology* that recovers energy stored in the capacitance of the column lines to reduce power consumption further.

The FPD33684 is designed for use in systems using dot inversion as the method of polarity inversion. Column inversion and N-line inversion are also supported. Other modes of polarity inversion including line inversion and frame inversion are not supported.

Digital video data inputs to the FPD33684 are received via a low power, low EMI Reduced Swing Differential Signaling (RSDS[™]) bus. The RSDS[™] digital video commands one of

64 gray level voltages on each output. Output voltages are driven with individual high drive, low offset operational amplifiers. Data loading and line buffering is accomplished by means of an internal, bi-directional shift register. FPD33684

GAMMA CORRECTION

The FPD33684 is designed to offer compatibility with a wide range of panel gamma characteristics. The output voltage is controlled by the digital data on the RSDSTM bus. Two identical R-DACs are used to program the output voltages. One R-DAC provides the high-polarity output voltages (voltages higher than V_{com}) and the other provides the low-polarity output voltages (voltages lower than V_{com}).

The FPD33684 is available with two R-DAC resistance curve options, both of which have been carefully designed to accurately match the natural, inverse gamma of a twisted nematic (TN) display with a 2.2 gamma transfer characteristic. A typical TN display, when operated with the FPD33684 drivers will produce a luminance with grayscale characteristics typical of CRT monitors. The R-DAC resistance values for the FPD33684A are shown in *Figure 3* and *Figure 4*. The

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R-DAC resistance values for the FPD33684B (designed to match the gamma curve of the Samsung S6C0666) are shown in *Figure 5* and *Figure 6*. Most applications will only need to provide references for each of the two ends of the two R-DACs (GMA1, GMA5, GMA6, and GMA10). Six additional, intermediate R-DAC tap points are available for further customization.

CHARGE CONSERVATION TECHNOLOGY

National Semiconductor's proprietary charge conservation technology significantly reduces power consumption. Charge conservation works by briefly switching all of the columns at the start of each line to a common node. This has the effect of redistributing the charge stored in the capacitance of the panel columns. Because half the columns are at voltages more positive than $V_{\rm com}$ and half are more negative, this redistribution of charge or "charge-sharing" has the effect of pulling all of the columns to a neutral voltage near the middle of the driver's dynamic range. Thus, the voltages on all the columns are driven approximately halfway toward their next value with no power expended. This dramatically reduces panel power dissipation (up to a theoretical limit of 50%) compared to conventional drivers which must drive each column through the entire voltage swing every time polarity is reversed.

'Smart' charge sharing is used to further optimize this feature. Data inversion is monitored and charge shared only across data ranges (when output polarity changes between adjacent lines). This is useful during n-line inversion when polarity changes do not occur at every line transition.

Charge sharing enables the FPD33684 to have faster output rise and fall times than drivers with convential amplifiers. This is due to the fact that the instantaneous currents supplied by the energy stored in the panel are much higher than the maximum output current of conventional drivers.

The CSTIME pin allows the user to set the duration of charge-sharing mode based on the panel capacitance and resistance. The length of charge-sharing is important because it must be long enough to allow all of the columns to equalize to the same value in order to achieve optimum power performance. The length of charge-mode is user programmable. There are two common methods to drive the CSTIME pin.

The first method is to actively drive the CSTIME input with a control signal. This may be achieved by connecting the LOAD signal to the CSTIME input. The width of the LOAD/CSTIME signal determines the amount of time spent in charge-sharing. This width may be optimized for a particular panel load. A 'typical' width is 800ns. If desired, the CSTIME pin may be driven independently, however, this will require an additional output from the timing controller.

At the rising edge of the CSTIME/LOAD input signal, the outputs enter charge-sharing mode. Outputs remain in charge-mode until the falling edge of the CSTIME/LOAD signal.

A second method for setting charge-time is to connect a resistor (R_{CSTIME}) and capacitor (C_{CSTIME}) in parallel between the CSTIME pin and ground. Only one resistor and capacitor is required for the entire display. At the rising edge of the LOAD signal, the CSTIME pin is internally pulled to V_{DD1} and then released (i.e. floated). At this time the outputs enter charge-sharing mode. The voltage on the CSTIME pin,

 $V_{\rm CSTIME},$ will then decay toward GND at a rate determined by the $R_{\rm CSTIME}$ and $C_{\rm CSTIME}$ time constant. When $V_{\rm CSTIME}$ reaches $V_{\rm DD1}/2$ the output mode switches from charge sharing to conventional amplifier drive mode. The charge-share mode time can be calculated using the following equation:

t_{charge-share}= 0.69 x R_{CSTIME} x C_{CSTIME}

RSDS™ DATA CHANNEL

The RSDS[™] data bus is comprised of nine differential data pairs and a differential clock. The nine channels are organized as three busses of three channels each. Each three channel bus corresponds on one of the three video colors, red, green and blue. Because the clocking is dual edged, the even fields of the 6-bit word are transmitted-received on a first clock and are followed by the odd fields. One full pixel (red, green, and blue subpixels) is transmitted every full pixelclock cycle.

OPTIONAL LINE BUFFERS

The FPD33684 provides two general purpose, unity gain output buffers, one located at each end of the input bank of the die. These buffers may be used to repair an open column line. The drive signal from the output of the faulted line can be stitched to the input of the repair buffer during the repair process. The output of the repair buffer is then routed to the other side of the column line making it possible to maintain fast rise and fall times on both ends of the afflicted column line.

PIN DESCRIPTIONS

The pin order configuration for the FPD33684 is shown in *Figure 7*. Optional pins do not need to be carried off a custom TCP or COF package but may require a connection to a neighboring pad on the die by a tie on the tape.

CLKP and CLKN—DATA CLOCK (INPUT)

Differential clock input for RSDS[™] data loading.

D00P-D22N - RSDS™ DATA BUS (INPUT)

D0xP-D0xN-Data for OUTPUTS 1,4,7...382 (red)

D1xP-D1xN-Data for OUTPUTS 2,5,8...383 (green)

D2xP-D2xN-Data for OUTPUTS 3,6,9...384 (blue)

Where x = 0 (LSB), 1 or 2 (MSB).

ENIO1/ENIO2 — DATA LOADING ENABLE 1 AND 2 (I/O)

The ENIO1/ ENIO2 pins are used to daisy chain the FPD33684 together with other FPD33684s. The first input in the chain is normally connected to the SP signal (or it's equivalent) on the timing controller. If UP = H, then the ENIO1 pin is configured as an input and the ENIO2 pin is configured as an output. If UP = L, then the ENIO2 pin is configured as an input and the ENIO1 pin is configured as an output.

INVERT --- DIGITAL DATA INVERT (INPUT)

When INVERT = H, RSDS data is inverted. The INVERT pin can be tied either high or low through connection to a neighboring pin, eliminating the need to bring the pin off the package.

LOAD — DATA LOAD (INPUT)

The rising edge of LOAD copies the digital video buffered by the shift register into a second latch for conversion to analog. The outputs are forced into charge share mode while load is high. When CSTIME = LOAD the falling edge ends the charge share time and the newly converted analog voltages are driven by the outputs.

POL—POLARITY (INPUT)

When POL = L, odd numbered outputs (1, 3, 5, ...383) are controlled by VGMA6 through VGMA10 and even numbered outputs are controlled by VGMA1 through VGMA5. When POL = H, odd numbered outputs are controlled by VGMA1 through VGMA5 and even numbered outputs are controlled by VGMA6 through VGMA10.

RPI1/ RPI2-REPAIR INPUT 1 AND 2 (INPUT)

The input signal for the repair line buffers. These buffers are optional and when not used, the input should be tied to ground. The pin can be tied to ground through a local pin on the TCP, eliminating the need to bring the repair amp inputs or outputs off the TCP.

RPO1/ RPO2—REPAIR OUTPUT 1 AND 2 (OUTPUT)

The output of the repair line buffers. These outputs are current buffered copies of their respective inputs.

UP -- DATA SHIFT DIRECTION -- UP OR DOWN (INPUT)

The UP pin controls the data shift direction. If UP is high then data is shifted "up" from output 1 to output 384, ENIO1 is configured as an input, and ENIO2 is an output. If UP is low then data is shifted "down" from output 384 to output 1, ENIO2 is an input, and ENIO1 is an output. The UP pin can be tied either high or low through connection to a neighboring pin, eliminating the need to bring the pin off the package.

CSTIME - CHARGE SHARE TIME

The CSTIME pin allows the user to set the duration of charge-sharing mode based on the panel capacitance and resistance.

V_{DD1}—DIGITAL VOLTAGE SUPPLY

Positive supply voltage for the digital logic functions of the driver.

V_{DD2}—ANALOG VOLTAGE SUPPLY

Positive supply voltage for the analog functions of the driver. $V_{GMA1}-V_{GMA10}$ —RDAC REFERENCES (INPUTS)

The reference voltages to the upper and lower RDACs used to control the inverse gamma transfer function of the driver.

Option - Any or all of the inputs V_{GMA2} through V_{GMA4} and V_{GMA7} through V_{GMA9} can be left undriven (floating).

V_{HBIAS}—HIGH BIAS CURRENT VOLTAGE SUPPLY

Optional positive supply voltage that provides a constant bias current to the output amplifiers to extend dynamic range. When separately provided, $V_{\rm HBIAS}$ must be 1.5V greater than $V_{\rm GMA1}$. When not separately provided, $V_{\rm HBIAS}$ must be tied to $V_{\rm DD2}$. In this configuration, $V_{\rm GMA1}$ must be held at or below $V_{\rm DD2}$ –1.5V.

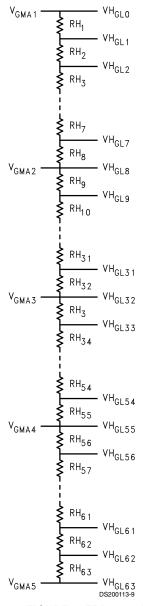
V_{ss1}—DIGITAL GROUND

Digital ground reference voltage.

V_{ss2}—ANALOG GROUND

Analog ground reference voltage.

Functional Description (Continued)



RH1	RDAC x 123/1008
RH2	RDAC x 69/1008
RH3	RDAC x 49/1008
RH4	RDAC x 42/1008
RH5	RDAC x 35/1008
RH6	RDAC x 28/1008
RH7	RDAC x 28/1008
RH8	RDAC x 21/1008
RH9	RDAC x 21/1008
RH10	RDAC x 14/1008
RH11	RDAC x 14/1008
RH12	RDAC x 10/1008
RH13	RDAC x 10/1008
RH14	RDAC x 9/1008
RH15	RDAC x 9/1008
RH16	RDAC x 8/1008
RH17	RDAC x 8/1008
RH18	RDAC x 8/1008
RH19	RDAC x 8/1008
RH20	RDAC x 8/1008
RH21	RDAC x 7/1008
RH22	RDAC x 7/1008
RH23	RDAC x 7/1008
RH24	RDAC x 7/1008
RH25	RDAC x 7/1008
RH26	RDAC x 7/1008
RH27	RDAC x 7/1008
RH28	RDAC x 7/1008
RH29	RDAC x 7/1008
RH30	RDAC x 7/1008
RH31	RDAC x 7/1008

	RH32	RDAC x 7/1008
1	RH33	RDAC x 7/1008
1	RH34	RDAC x 7/1008
1	RH35	RDAC x 7/1008
1	RH36	RDAC x 7/1008
1	RH37	RDAC x 7/1008
1	RH38	RDAC x 7/1008
1	RH39	RDAC x 7/1008
1	RH40	RDAC x 7/1008
1	RH41	RDAC x 7/1008
1	RH42	RDAC x 7/1008
1	RH43	RDAC x 7/1008
1	RH44	RDAC x 8/1008
1	RH45	RDAC x 8/1008
1	RH46	RDAC x 8/1008
1	RH47	RDAC x 8/1008
1	RH48	RDAC x 8/1008
1	RH49	RDAC x 8/1008
1	RH50	RDAC x 8/1008
1	RH51	RDAC x 9/1008
1	RH52	RDAC x 9/1008
1	RH53	RDAC x 10/1008
1	RH54	RDAC x 10/1008
1	RH55	RDAC x 10/1008
1	RH56	RDAC x 10/1008
1	RH57	RDAC x 13/1008
1	RH58	RDAC x 15/1008
1	RH59	RDAC x 17/1008
1	RH60	RDAC x 21/1008
	RH61	RDAC x 35/1008
1	RH62	RDAC x 48/1008
_	RH63	RDAC x 62/1008

FIGURE 3. FPD33684A R-DAC Transfer Characteristic (continued in next figure)

RL32

RL33

RL34

RL35

RL36

RL37

RL38

RL39

RL40

RL41

RL42

RL43

RL44

RL45

RL46

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RL48

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RL58

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RL63

RDAC x 7/1008

RDAC x 7/1008 RDAC x 7/1008

RDAC x 7/1008

RDAC x 7/1008

RDAC x 7/1008

RDAC x 7/1008

RDAC x 7/1008

RDAC x 7/1008

RDAC x 7/1008

RDAC x 7/1008

RDAC x 8/1008

RDAC x 9/1008

RDAC x 9/1008

RDAC x 10/1008

RDAC x 10/1008

RDAC x 10/1008

RDAC x 10/1008

RDAC x 13/1008

RDAC x 15/1008

RDAC x 17/1008

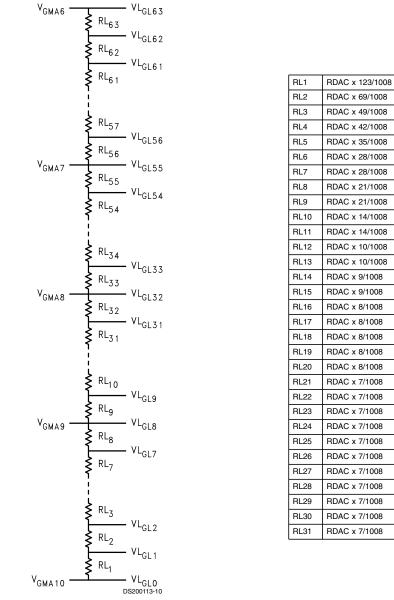
RDAC x 21/1008

RDAC x 35/1008

RDAC x 48/1008

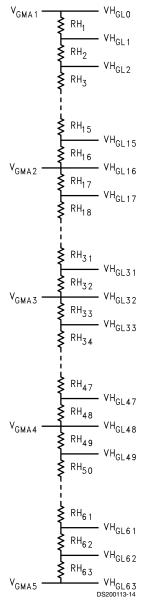
RDAC x 62/1008

RDAC 7/1008





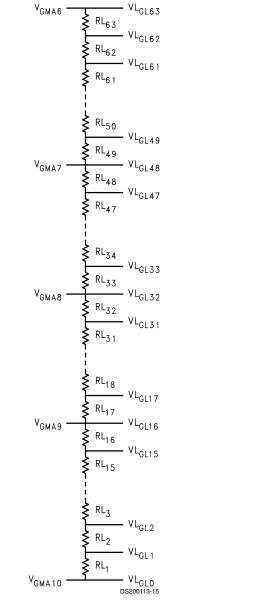
Functional Description (Continued)



RH1	RDAC x 27/1008
RH2	RDAC x 27/1008
RH3	RDAC x 27/1008
RH4	RDAC x 27/1008
RH5	RDAC x 27/1008
RH6	RDAC x 27/1008
RH7	RDAC x 27/1008
RH8	RDAC x 27/1008
RH9	RDAC x 27/1008
RH10	RDAC x 27/1008
RH11	RDAC x 27/1008
RH12	RDAC x 27/1008
RH13	RDAC x 24/1008
RH14	RDAC x 24/1008
RH15	RDAC x 21/1008
RH16	RDAC x 20/1008
RH17	RDAC x 18/1008
RH18	RDAC x 18/1008
RH19	RDAC x 18/1008
RH20	RDAC x 17/1008
RH21	RDAC x 16/1008
RH22	RDAC x 15/1008
RH23	RDAC x 14/1008
RH24	RDAC x 14/1008
RH25	RDAC x 13/10078
RH26	RDAC x 13/1008
RH27	RDAC x 12/1008
RH28	RDAC x 12008
RH29	RDAC x 11/1008
RH30	RDAC x 11/1008
RH31	RDAC x 10/1008

RH32	RDAC x 10/1008
RH33	RDAC x 9/1008
RH34	RDAC x 9/1008
RH35	RDAC x 9/1008
RH36	RDAC x 9/1008
RH37	RDAC x 9/1008
RH38	RDAC x 9/1008
RH39	RDAC x 9/1008
RH40	RDAC x 9/1008
RH41	RDAC x 9/1008
RH42	RDAC x 9/1008
RH43	RDAC x 9/1008
RH44	RDAC x 9/1008
RH45	RDAC x 9/1008
RH46	RDAC x 9/1008
RH47	RDAC x 10/1008
RH48	RDAC x 10/1008
RH49	RDAC x 11/1008
RH50	RDAC x 12/1008
RH51	RDAC x 12/1008
RH52	RDAC x 13/1008
RH53	RDAC x 13/1008
RH54	RDAC x 14/1008
RH55	RDAC x 14/1008
RH56	RDAC x 16/1008
RH57	RDAC x 16/1008
RH58	RDAC x 17/1008
RH59	RDAC x 17/1008
RH60	RDAC x 18/1008
RH61	RDAC x 18/1008
RH62	RDAC x 18/1008
RH63	RDAC x 18/1008

FIGURE 5. FPD33684B R-DAC Transfer Characteristic (matches Samsung S6C0666) (continued in next figure)



RL1	RDAC x 27/1008	RL32	RDAC x 10/1008
RL2	RDAC x 27/1008	RL33	RDAC x 9/1008
RL3	RDAC x 27/1008	RL34	RDAC x 9/1008
RL4	RDAC x 27/1008	RL35	RDAC x 9/1008
RL5	RDAC x 27/1008	RL36	RDAC x 9/1008
RL6	RDAC x 27/1008	RL37	RDAC x 9/1008
RL7	RDAC x 27/1008	RL38	RDAC x 9/1008
RL8	RDAC x 27/1008	RL39	RDAC x 9/1008
RL9	RDAC x 27/1008	RL40	RDAC x 9/1008
RL10	RDAC x 27/1008	RL41	RDAC x 9/1008
RL11	RDAC x 27/1008	RL42	RDAC x 9/1008
RL12	RDAC x 27/1008	RL43	RDAC x 9/1008
RL13	RDAC x 24/1008	RL44	RDAC x 9/1008
RL14	RDAC x 24/1008	RL45	RDAC x 9/1008
RL15	RDAC x 21/1008	RL46	RDAC x 9/1008
RL16	RDAC x 20/1008	RL47	RDAC x 10/1008
RL17	RDAC x 18/1008	RL48	RDAC x 10/1008
RL18	RDAC x 18/1008	RL49	RDAC x 11/1008
RL19	RDAC x 18/1008	RL50	RDAC x 12/1008
RL20	RDAC x 17/1008	RL51	RDAC x 12/1008
RL21	RDAC x 16/1008	RL52	RDAC x 13/1008
RL22	RDAC x 15/1008	RL53	RDAC x 13/1008
RL23	RDAC x 14/1008	RL54	RDAC x 14/1008
RL24	RDAC x 14/1008	RL55	RDAC x 14/1008
RL25	RDAC x 13/1008	RL56	RDAC x 16/1008
RL26	RDAC x 13/1008	RL57	RDAC x 16/1008
RL27	RDAC x 12/1008	RL58	RDAC x 17/1008
RL28	RDAC x 12/1008	RL59	RDAC x 17/1008
RL29	RDAC x 11/1008	RL60	RDAC x 18/1008
RL30	RDAC x 11/1008	RL61	RDAC x 18/1008
RL31	RDAC x 10/1008	RL62	RDAC x 18/1008
		RL63	RDAC x 18/1008

FIGURE 6. FPD33684B R-DAC Transfer Characteristic (matches Samsung S6C0666) (conitnued from previous figure)

Functional Description (Continued)

INPUTS			OUTPUTS
RPI2]		384
RPIO2	1		383
VSS2	1		382
VDD2	1		381
ENIO2	1		
D22P	1		
D22N	1		
D21P	1		
D21N	1		
D20P	1		
D20N	1		
D12P	1		
D12N	1		
D11P	1		
D11N	1		
D10P	1		
D10N	1		
VCS	1		
VDD1	1		
UP	1		
VGMA10	1		
VGMA9	1		
VGMA8	1		
VGMA7	1	84	
VGMA6	1	336	
VHBIAS	1	EPD	
VGMA5	1		
VGMA4	1		
VGMA3	1		
VGMA2	1		
VGMA1	1		
VSS1	1		
CLKP	1		
CLKN	1		
LOAD	1		
CSTIME	1		
POL	1		
INVERT	1		
D02P	1		
D02N	1		
D01P	1		
D01N	1		
D00P	1		
D00N	1		6
ENIO1	1		5
VSS2	1		4
VDD2	1		3
	1		
RPO1			2
	RPI2RPI02VSS2VDD2ENI02D22PD21PD21ND20PD20ND12PD11PD11PD11ND10PVCSVDD1UPVGMA10VGMA5VGMA5VGMA4VGMA1VSS1CLKPCLKPCLKNLOADD01PD01NVGMA1VGMA2VGMA2VGMA1VSS1CLKPCLKNLOADCSTIMEPOLINVERTD02PD01ND01PD01ND00PD00NENI01VSS2VDD2	RPI2 RPI02 VSS2 VDD2 ENI02 D22P D21P D21P D20N D12P D12N D11P D10N VCS VDD1 UP VGMA10 VGMA3 VGMA5 VGMA5 VGMA1 VSS1 CLKP CLKN LOAD VGMA1 VSS1 CLKP CLKN LOAD VGMA1 VSS1 CLKN LOAD CSTIME POL D01P D02N D01P D02N D01N D00P D01N D00P D01N D00P D00N ENI01 VSS2 VDD2	RPI2 RPI02 VSS2 VDD2 ENI02 D22P D22N D21P D21N D20N D12P D11N D11P D10N VCS VDD1 UP VGMA10 VGMA5 VGMA5 VGMA2 VGMA1 VSS1 CLKP CLKP CLKN LOAD CSTIME POL NUVERT D02P D01N VSS2 VD01 VSS1 CLKN LOAD CSTIME POL D01P D01P D02N D01P D01P D00P D00N ENI01 VSS2 VDD2

Note: This figure represents a FPD33684 die oriented pad side up. FIGURE 7. FPD33684 I/O Configuration

Packaging

The FPD33684 is available in TCP or as singulated die.

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